

Technical documentation



Support & ക training



DLP480RE DLPS160A - APRIL 2019 - REVISED JUNE 2022

DLP480RE 0.48 WUXGA DMD

1 Features

- 0.48-inch diagonal micromirror array
 - WUXGA (1920 × 1200) display resolution
 - 5.4 Micron micromirror pitch
 - ±17° micromirror tilt (relative to flat surface) Bottom illumination
- 2xLVDS input data bus
- Dedicated DLPC4422 display controller, DLPA100 ٠ power management IC and motor driver for reliable operation

2 Applications

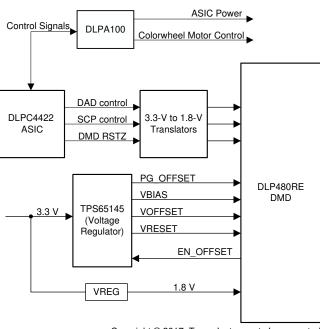
- WUXGA display
- Smart display
- **Digital signage**
- **Business projector** •
- Education projector •

3 Description

The TI DLP480RE digital micromirror device (DMD) is a digitally controlled micro-electromechanical system (MEMS) spatial light modulator (SLM) that enables high resolution WUXGA display systems. The DLP480RE DMD, together with the DLPC4422 display controller and DLPA100 power and motor driver, comprise the DLP® 0.48-inch WUXGA chipset. This chipset serves as an optimal solution for any system that demands a cost-effective, high-resolution display.

Device Information

PART NUMBER	PACKAGE	BODY SIZE (NOM)
DLP480RE	FXG (257)	32 mm × 22 mm



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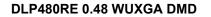






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4 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

C	hanges from Revision * (April 2019) to Revision A (June 2022)	Page
•	This document is updated per the latest Texas Instruments and industry data sheet standards	1
•	Updated Timing Requirements	15
•	Updated Figure 6-3	15



5 Pin Configuration and Functions

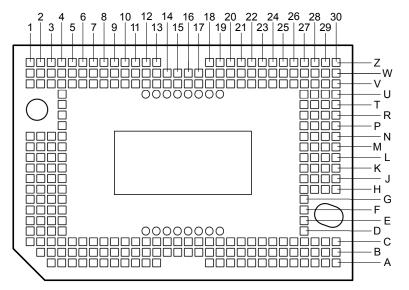


Figure 5-1. Series 410 257-pin FXG Bottom View

CAUTION

To ensure reliable, long-term operation of the .48-inch WUXGA S410 DMD, it is critical to properly manage the layout and operation of the signals identified in the table below. For specific details and guidelines, refer to the *PCB Design Requirements for TI DLP Standard TRP Digital Micromirror Devices* application report before designing the board.

DIN							TRACE
PIN		I/O ⁽³⁾	SIGNAL	DATA	INTERNAL	DESCRIPTION	TRACE LENGTH
NAME	NO.	1/0(**	SIGNAL	RATE TERMINATION DESCRIPTION		(mil)	
D_AN(0)	C6						
D_AN(1)	C3						
D_AN(2)	E1						
D_AN(3)	C4						
D_AN(4)	D1	1					
D_AN(5)	B8]					
D_AN(6)	F4	1					
D_AN(7)	E3			DDR	Differential	No connect	905.0
D_AN(8)	C11	- NC	LVDS	DDR	Differential	No connect	805.0
D_AN(9)	F3	1					
D_AN(10)	K4						
D_AN(11)	H3						
D_AN(12)	J3	1					
D_AN(13)	C13	1					
D_AN(14)	A5	1					
D_AN(15)	A3						

Table 5-1. Pin Functions

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			Table	5-1. Pi	n Functions (continued)	
PIN		I/O ⁽³⁾	SIGNAL	DATA	INTERNAL	DESCRIPTION	TRACE LENGTH
NAME	NO.		SIGNAL	RATE	TERMINATION	DESCRIPTION	(mil)
D_AP(0)	C7						
D_AP(1)	C2						
D_AP(2)	E2						
D_AP(3)	B4						
D_AP(4)	C1						
D_AP(5)	B7						
D_AP(6)	E4						
D_AP(7)	D3	NC	LVDS	DDR	Differential	No connect	805.0
D_AP(8)	C12		LVDS	DDR	Differentia		005.0
D_AP(9)	F2						
D_AP(10)	J4						
D_AP(11)	G3						
D_AP(12)	J2						
D_AP(13)	C14						
D_AP(14)	A6						
D_AP(15)	A4						
D_BN(0)	N4						
D_BN(1)	Z11						
D_BN(2)	W4						
D_BN(3)	W10						805.0
D_BN(4)	L1						
D_BN(5)	V8						
D_BN(6)	W6						
D_BN(7)	M1	NC	LVDS	DDR	Differential		
D_BN(8)	R4		LVDS	DDR	Differential	No connect	005.0
D_BN(9)	W1						
D_BN(10)	U4						
D_BN(11)	V2						
D_BN(12)	Z5						
D_BN(13)	N3						
D_BN(14)	Z2]					
D_BN(15)	L4						

Table 5-1. Pin Functions (continued)



PIN				DATA			TRACE
NAME	NO.	I/O ⁽³⁾	SIGNAL	RATE	TERMINATION	DESCRIPTION	LENGTH (mil)
D_BP(0)	M4						
D_BP(1)	Z12	1					
D_BP(2)	Z4	1					
D_BP(3)	Z10						
D_BP(4)	L2						
D_BP(5)	V9						
D_BP(6)	W7						
D_BP(7)	N1	NC			Differential		005.0
D_BP(8)	P4	NC	LVDS	DDR	Differential	No connect	805.0
D_BP(9)	V1						
D_BP(10)	T4	1					
D_BP(11)	V3	1					
D_BP(12)	Z6	1					
D_BP(13)	N2						
D_BP(14)	Z3						
D_BP(15)	L3						
D_CN(0)	H27						
D_CN(1)	A20						
D_CN(2)	H28						
D_CN(3)	K28						005.0
D_CN(4)	K30						
D_CN(5)	C23						
D_CN(6)	G27	1					
D_CN(7)	J30		LVDS	ססס	Differential		
D_CN(8)	B24	- 1	LVDS	DDR	Differential	Data negative	805.0
D_CN(9)	A21						
D_CN(10)	A27						
D_CN(11)	C29]					
D_CN(12)	A26	1					
D_CN(13)	C25	1					
D_CN(14)	A29	1					
D_CN(15)	C30]					

Table 5-1. Pin Functions (continued)

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DIN			Iadie		n Functions (TDAOE
PIN		I/O ⁽³⁾	3) SIGNAL DATA INTERNAL	DESCRIPTION	TRACE LENGTH		
NAME	NO.			RATE	TERMINATION		(mil)
D_CP(0)	J27						
D_CP(1)	A19						
D_CP(2)	H29						
D_CP(3)	K27						
D_CP(4)	K29						
D_CP(5)	C22						
D_CP(6)	F27						
D_CP(7)	H30	1	LVDS	DDR	Differential	Data positive	805.0
D_CP(8)	B25		LVDS	DDR	Differential		805.0
D_CP(9)	B21						
D_CP(10)	B27						
D_CP(11)	C28						
D_CP(12)	A25						
D_CP(13)	C24						
D_CP(14)	A28						
D_CP(15)	B30						
D_DN(0)	V25						805.0
D_DN(1)	V28						
D_DN(2)	T30						
D_DN(3)	V27						
D_DN(4)	U30						
D_DN(5)	W23						
D_DN(6)	R27	1					
D_DN(7)	T28	- 1	LVDS	DDR	Differential		
D_DN(8)	V20			DUK	Differential	Data negative	
D_DN(9)	R28						
D_DN(10)	L27]					
D_DN(11)	N28]					
D_DN(12)	M28						
D_DN(13)	V18						
D_DN(14)	Z26						
D_DN(15)	Z28	1					

Table 5-1. Pin Functions (continued)



Table 5-1. Pin Functions (continued) TRACE PIN DATA INTERNAL I/O⁽³⁾ SIGNAL DESCRIPTION LENGTH RATE TERMINATION NAME NO. (mil) D DP(0) V24 805.0 D DP(1) V29 D_DP(2) T29 D_DP(3) W27 D DP(4) V30 D_DP(5) W24 D DP(6) T27 D DP(7) U28 LVDS Differential L DDR Data positive D DP(8) V19 D_DP(9) R29 D_DP(10) M27 D DP(11) P28 D_DP(12) M29 D DP(13) V17 D_DP(14) Z25 D DP(15) Z27 SCTRL_AN G1 F1 SCTRL AP NC LVDS 805.0 DDR Differential No connect SCTRL BN V5 SCTRL_BP V4 C26 LVDS DDR Differential 805.0 SCTRL CN Serial control negative Т SCTRL CP C27 T LVDS DDR Differential Serial control positive 805.0 SCTRL DN P30 L LVDS DDR Differential Serial control negative 805.0 SCTRL DP R30 LVDS DDR Differential Serial control positive 805.0 L DCLK_AN H2 DCLK AP H1 NC LVDS Differential No connect 805.0 DCLK BN V6 DCLK_BP V7 LVDS 805.0 DCLK CN D27 L Differential Clock negative DCLK CP E27 I LVDS Differential Clock positive 805.0 N29 L LVDS Differential 805.0 DCLK_DN Clock negative 805.0 DCLK_DP N30 LVDS Differential Clock positive I Serial communications port clock. Active only SCPCLK A10 L LVCMOS Pulldown when SCPENZ is logic low Serial communications port Data Input. SCPDI L LVCMOS SDR Pulldown A12 Synchronous to SCPCLK rising edge SCPENZ C10 Pulldown L LVCMOS Serial communications port enable active low SCPDO A11 0 LVCMOS SDR Serial communications port output RESET ADDR(0) Z13 W13 RESET ADDR(1) Т LVCMOS Pulldown Reset driver address select RESET ADDR(2) V10 W14 RESET_ADDR(3) RESET_MODE(0) W9 Reset driver mode select LVCMOS Pulldown RESET_SEL(0) V14 L Reset driver level select RESET_SEL(1) Z8 Reset driver level select

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PIN				DATA INTERNAL			TRACE
NAME	NO.	I/O ⁽³⁾	SIGNAL	RATE	TERMINATION	DESCRIPTION	LENGTH (mil)
RESET_STROBE	Z9	I	LVCMOS		Pulldown	Rising edge latches in RESET_ADDR, RESET_MODE, and RESET_SEL	
PWRDNZ	A8	I	LVCMOS		Pulldown	Active low device reset	
RESET_OEZ	W15	I	LVCMOS		Pullup	Active low output enable for internal reset driver circuits	
RESET_IRQZ	V16	0	LVCMOS			Active low output interrupt to DLP [®] display controller	
EN_OFFSET	C9	0	LVCMOS			Active high enable for external VOFFSET regulator	
PG_OFFSET	A9	I	LVCMOS		Pullup	Active low fault from external VOFFSET regulator	
TEMP_N	B18		Analog			Temperature sensor diode cathode	
TEMP_P	B17		Analog			Temperature sensor diode anode	
RESERVED	D12, D13, D14, D15, D16, D17, D18, D19, U12, U13, U14, U15	NC	Analog		Pulldown	Do not connect on the printed circuit board (PCB). No connect. No electrical connections from CMOS bond pad to package pin	
No Connect	U16, U17, U18, U19	NC				No connect. No electrical connection from CMOS bond pad to package pin	
RESERVED_BA	W11						
RESERVED_BB	B11	0	LVCMOS			Do not connect on the printed circuit board	
RESERVED_BC	Z20		LVONIOO			(PCB)	
RESERVED_BD	C18						
RESERVED_PFE	A18	. I	LVCMOS		Pulldown	Connect to ground on the printed circuit board	
RESERVED_TM	C8					(PCB)	
RESERVED_TP0	Z19					Do not connect on the printed circuit board	
RESERVED_TP1	W20	I	Analog			(PCB)	
RESERVED_TP2	W19						
VBIAS ⁽¹⁾	C15, C16, V11, V12	Р	Analog			Supply voltage for positive bias level of micromirror reset signal	
VRESET ⁽¹⁾	G4, H4, J1, K1	Р	Analog			Supply voltage for negative reset level of micromirror reset signal	
VOFFSET ⁽¹⁾	A30, B2, M30, Z1, Z30	Р	Analog			Supply voltage for HVCMOS logic. Supply voltage for positive offset level of micromirror reset signal. Supply voltage for stepped high voltage at micromirror address electrodes	
VCC ⁽¹⁾	A24, A7, B10, B13, B16, B19, B22, B28, B5, C17, C20, D4, J29, K2, L29, M2, N27, U27, V13, V15, V22, W17, W21, W26, W29, W3, Z18, Z23, Z29, Z7	Ρ	Analog			Supply voltage for LVCMOS core. Supply voltage for positive offset level of micromirror reset signal during Power down. Supply voltage for normal high level at micromirror address electrodes	



					· · · · · /		
PIN			DATA	INTERNAL		TRACE	
NAME	NO.	I/O ⁽³⁾	SIGNAL		TERMINATION	DESCRIPTION	LENGTH (mil)
VSS ⁽²⁾	A13, A22, A23, B12, B14, B15, B20, B23, B26, B29, B3, B6, B9, C19, C21, C5, D2, G2, J28, K3, L28, L30, M3, P27, P29, U29, V21, V23, V26, W12, W16, W18, W2, W25, W28, W30, W5, W8, Z21, Z22, Z24	G				Device ground. Common return for all power	

Table 5-1. Pin Functions (continued)

V_{BIAS}, V_{CC}, V_{OFFSET}, and V_{RESET} power supplies must be connected for proper DMD operation.
 V_{SS} must be connected for proper DMD operation.
 I = Input, O = Output, P = Power, G = Ground, NC = No connect



6 Specifications

6.1 Absolute Maximum Ratings

Over operating free-air temperature range (unless otherwise noted)⁽¹⁾

		MIN	MAX	UNIT
SUPPLY VOLTAGE	S			
V _{CC}	Supply voltage for LVCMOS core logic ⁽²⁾	-0.5	2.3	V
V _{OFFSET}	Supply voltage for HVCMOS and micromirror electrode ^{(2) (3)}	-0.5	11	V
V _{BIAS}	Supply voltage for micromirror electrode ⁽²⁾	-0.5	19	V
V _{RESET}	Supply voltage for micromirror electrode ⁽²⁾	-15	-0.3	V
V _{BIAS} – V _{OFFSET}	Supply voltage difference (absolute value) ⁽⁴⁾		11	V
V _{BIAS} – V _{RESET}	Supply voltage difference (absolute value) ⁽⁵⁾		34	V
INPUT VOLTAGES				
	Input voltage for all other LVCMOS input pins ⁽²⁾	-0.5	V _{CC} + 0.5	V
V _{ID}	Input differential voltage (absolute value) ⁽⁶⁾		500	mV
I _{ID}	Input differential current ⁽⁷⁾		6.3	mA
Clocks				
fсlock	Clock frequency for LVDS interface, DCLK_C		400	MHz
<i>f</i> сlocк	Clock frequency for LVDS interface, DCLK_D		400	MHz
ENVIRONMENTAL		1		
T _{ARRAY} and	Temperature, operating ⁽⁸⁾	0	90	°C
T _{WINDOW}	Temperature, non-operating ⁽⁸⁾	-40	90	°C
T _{delta}	Absolute Temperature difference between any point on the window edge and the ceramic test point TP1 ⁽⁹⁾		30	°C
T _{DP}	Dew Point Temperature, operating and non-operating (noncondensing)		81	°C

(1) Operation outside the Absolute Maximum Ratings may cause permanent device damage. Absolute Maximum Ratings do not imply functional operation of the device at these or any other conditions beyond those listed under Recommended Operating Conditions. If outside the Recommended Operating Conditions but within the Absolute Maximum Ratings, the device may not be fully functional, and this may affect device reliability, functionality, performance, and shorten the device lifetime.

(2) All voltages are referenced to common ground V_{SS}. V_{BIAS}, V_{CC}, V_{OFFSET}, and V_{RESET} power supplies are all required for proper DMD operation. V_{SS} must also be connected.

- (3) V_{OFFSET} supply transients must fall within specified voltages.
- (4) Exceeding the recommended allowable voltage difference between V_{BIAS} and V_{OFFSET} may result in excessive current draw.
- (5) Exceeding the recommended allowable voltage difference between V_{BIAS} and V_{RESET} may result in excessive current draw.

(6) This maximum LVDS input voltage rating applies when each input of a differential pair is at the same voltage potential.

(7) LVDS differential inputs must not exceed the specified limit or damage may result to the internal termination resistors.

(8) The highest temperature of the active array (as calculated using Section 7.6) or of any point along the window edge as defined in Figure 7-2. The locations of thermal test points TP2, TP3, TP4 and TP5 in Figure 7-2 are intended to measure the highest window edge temperature. If a particular application causes another point on the window edge to be at a higher temperature, use that point.

(9) Temperature difference is the highest difference between the ceramic test point 1 (TP1) and anywhere on the window edge as shown in Figure 7-2. The window test points TP2, TP3, TP4 and TP5 shown in Figure 7-2 are intended to result in the worst case difference. If a particular application causes another point on the window edge to result in a larger difference temperature, use that point.

6.2 Storage Conditions

Applicable for the DMD as a component or non-operating in a system

		MIN	MAX	UNIT
T _{stg}	DMD storage temperature	-40	80	°C
T _{DP-AVG}	Average dew point temperature, (non-condensing) ⁽¹⁾		28	°C
T _{DP-ELR}	Elevated dew point temperature range , (non-condensing) (2)	28	36	°C
CT _{ELR}	Cumulative time in elevated dew point temperature range		24	Months

(1) The average over time (including storage and operating) that the device is not in the elevated dew point temperature range.

(2) Limit exposure to dew point temperatures in the elevated range during storage and operation to less than a total cumulative time of CT_{ELR}.



6.3 ESD Ratings

			VALUE	UNIT
V	Electrostatic discharge	Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 ⁽¹⁾	±2000	V
V _(ESD) Electrostatic discharge	Charged device model (CDM), per JEDEC specification JESD22-C101 ⁽²⁾	±500	v	

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

6.4 Recommended Operating Conditions

Over operating free-air temperature range (unless otherwise noted). The functional performance of the device specified in this data sheet is achieved when operating the device within the limits defined by this table. No level of performance is implied when operating the device above or below these limits.

		MIN	NOM	MAX	UNIT
VOLTAGE SUPP	LY				
V _{CC}	LVCMOS logic supply voltage ⁽¹⁾	1.65	1.8	1.95	V
V _{OFFSET}	Mirror electrode and HVCMOS voltage ⁽¹⁾ ⁽²⁾	9.5	10	10.5	V
V _{BIAS}	Mirror electrode voltage ⁽¹⁾	17.5	18	18.5	V
V _{RESET}	Mirror electrode voltage ⁽¹⁾	-14.5	-14	-13.5	V
V _{BIAS} – V _{OFFSET}	Supply voltage difference (absolute value) ⁽³⁾			10.5	V
V _{BIAS} – V _{RESET}	Supply voltage difference (absolute value) ⁽⁴⁾			33	V
LVCMOS INTER	FACE				
V _{IH(DC)}	DC input high voltage ⁽⁵⁾	0.7 × V _{CC}	١	/ _{CC} + 0.3	V
V _{IL(DC)}	DC input low voltage ⁽⁵⁾	-0.3	C).3 × V _{CC}	V
V _{IH(AC)}	AC input high voltage ⁽⁵⁾	0.8 × V _{CC}	١	/ _{CC} + 0.3	V
V _{IL(AC)}	AC input low voltage ⁽⁵⁾	-0.3	C).2 × V _{CC}	V
t _{PWRDNZ}	PWRDNZ pulse duration ⁽⁶⁾	10			ns
SCP INTERFACE					
f _{scpclk}	SCP clock frequency ⁽⁷⁾			500	kHz
tscp_pd	Propagation delay, Clock to Q, from rising–edge of SCPCLK to valid SCPDO ⁽⁸⁾	0		900	ns
SCP_NEG_ENZ	Time between falling-edge of SCPENZ and the first rising- edge of SCPCLK	1			μs
SCP_POS_ENZ	Time between falling-edge of SCPCLK and the rising-edge of SCPENZ	1			μs
SCP_DS	SCPDI Clock setup time (before SCPCLK falling edge) ⁽⁸⁾	800			ns
tscp_dh	SCPDI Hold time (after SCPCLK falling edge) ⁽⁸⁾	900			ns
SCP_PW_ENZ	SCPENZ inactive pulse duration (high level)	2			μs



Over operating free-air temperature range (unless otherwise noted). The functional performance of the device specified in this data sheet is achieved when operating the device within the limits defined by this table. No level of performance is implied when operating the device above or below these limits.

		MIN	NOM	MAX	UNIT
LVDS INTERF	ACE				
fclock	Clock frequency for LVDS interface (all channels), DCLK ⁽⁹⁾			400	MHz
V _{ID}	Input differential voltage (absolute value) ⁽¹⁰⁾	150	300	440	mV
V _{CM}	Common mode voltage ⁽¹⁰⁾	1100	1200	1300	mV
V _{LVDS}	LVDS voltage ⁽¹⁰⁾	880		1520	mV
t _{LVDS_RSTZ}	Time required for LVDS receivers to recover from PWRDNZ			2000	ns
Z _{IN}	Internal differential termination resistance	80	100	120	Ω
Z _{LINE}	Line differential impedance (PWB/trace)	90	100	110	Ω
ENVIRONME	NTAL				
T _{ARRAY}	Array temperature, Long-term operational ⁽¹¹⁾ (12) (13) (14)	10		40 to 70 ⁽¹³⁾	°C
	Array temperature, Short–term operational ^{(12) (15)}	0		10	°C
T _{WINDOW}	Window temperature – operational ⁽¹⁹⁾ ⁽²¹⁾			85	°C
T _{DELTA}	Absolute temperature difference between any point on the window edge and the ceramic test point TP1 ⁽¹⁶⁾ ⁽¹⁷⁾			14	°C
T _{DP -AVG}	Average dew point temperature (non-condensing) ⁽¹⁸⁾			28	°C
T _{DP-ELR}	Elevated dew point temperature range (non-condensing) ⁽²⁰⁾	28		36	°C
CT _{ELR}	Cumulative time in elevated dew point temperature range			24	Months
L	Operating system luminance ⁽¹⁷⁾			4000	lm
ILL _{UV}	Illumination Wavelengths < 395 nm ⁽¹¹⁾		0.68	2.00	mW/cm ²
ILL _{VIS}	Illumination Wavelengths between 395 nm and 800 nm	Ther	mally limited		mW/cm ²
ILL _{IR}	Illumination Wavelengths > 800 nm			10	mW/cm ²
ILL _θ	Illumination Marginal Ray Angle ⁽²¹⁾			55	deg

(1) All voltages are referenced to common ground VSS. VBIAS, VCC, VOFFSET, and VRESET power supplies are all required for proper DMD operation. VSS must also be connected.

(2) VOFFSET supply transients must fall within specified max voltages.

- (3) To prevent excess current, the supply voltage difference |VBIAS VOFFSET| must be less than specified limit. See Section 9, Figure 9-1, and Table 9-1.
- (4) To prevent excess current, the supply voltage difference |VBIAS VRESET| must be less than specified limit. See Section 9, Figure 9-1, and Table 9-1.
- (5) Low-speed interface is LPSDR and adheres to the Electrical Characteristics and AC/DC Operating Conditions table in JEDEC Standard No. 209B, "Low-Power Double Data Rate (LPDDR)" JESD209B.Tester Conditions for VIH and VIL.
 - Frequency = 60 MHz. Maximum Rise Time = 2.5 ns @ (20% 80%)
 - Frequency = 60 MHz. Maximum Fall Time = 2.5 ns @ (80% 20%)
- (6) PWRDNZ input pin resets the SCP and disables the LVDS receivers. PWRDNZ input pin overrides SCPENZ input pin and tristates the SCPDO output pin.
- (7) The SCP clock is a gated clock. Duty cycle must be 50% ± 10%. SCP parameter is related to the frequency of DCLK.

(8) See Figure 6-2.

- (9) See LVDS Timing Requirements in Section 6.8 and Figure 6-6.
- (10) See Figure 6-5 LVDS Waveform Requirements.
- (11) Simultaneous exposure of the DMD to the maximum Section 6.4 for temperature and UV illumination reduces device lifetime.
- (12) The array temperature cannot be measured directly and must be computed analytically from the temperature measured at test point 1 (TP1) shown in Figure 7-2 and the package thermal resistance Section 7.6.
- (13) Per Figure 6-1, the maximum operational array temperature must be derated based on the micromirror landed duty cycle that the DMD experiences in the end application. See Section 7.7 for a definition of micromirror landed duty cycle.
- (14) Long-term is defined as the usable life of the device.
- (15) Array temperatures beyond those specified as long-term are recommended for short-term conditions only (power-up). Short-term is defined as cumulative time over the usable life of the device and is less than 500 hours.
- (16) Temperature difference is the highest difference between the ceramic test point 1 (TP1) and anywhere on the window edge as shown in Figure 7-2. The window test points TP2, TP3, TP4 and TP5 shown in Figure 7-2 are intended to result in the worst case difference temperature. If a particular application causes another point on the window edge to result in a larger difference in temperature, use that point.



- (17) DMD is qualified at the combination of the maximum temperature and maximum lumens specified. Operation of the DMD outside of these limits has not been tested.
- (18) The average over time (including storage and operating) that the device is not in the elevated dew point temperature range.
- (19) The locations of Thermal Test Points TP2, TP3, TP4 and TP5 in Figure 10 are intended to measure the highest window edge temperature. For most applications, the locations shown are representative of the highest window edge temperature. If a particular application causes additional points on the window edge to be at a higher temperature, add those test points
- (20) Limit exposure to dew point temperatures in the elevated range during storage and operation to less than a total cumulative time of CT_{ELR}.
- (21) The maximum marginal ray angle of the incoming illumination light at any point in the micromirror array, including Pond of Micromirrors (POM), should not exceed 55 degrees from the normal to the device array plane. The device window aperture has not necessarily been designed to allow incoming light at higher maximum angles to pass to the micromirrors, and the device performance has not been tested nor qualified at angles exceeding this. Illumination light exceeding this angle outside the micromirror array (including POM) will contribute to thermal limitations described in this document, and may negatively affect lifetime.

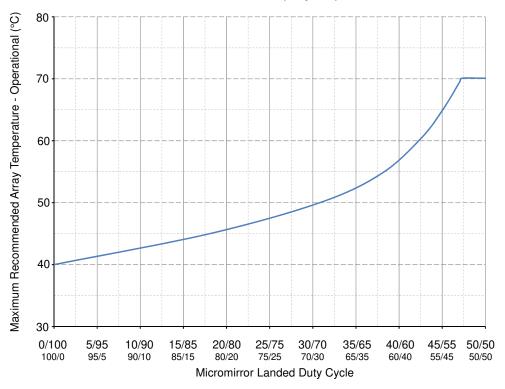


Figure 6-1. Maximum Recommended Array Temperature - Derating Curve

6.5 Thermal Information

	DLP480RE	
THERMAL METRIC	FXG Package	UNIT
	257 PINS	
Thermal resistance, active area to test point 1 (TP1) ⁽¹⁾	0.90	°C/W

(1) The DMD is designed to conduct absorbed and dissipated heat to the back of the package. The cooling system must be capable of maintaining the package within the temperature range specified in the *Recommended Operating Conditions*. The total heat load on the DMD is largely driven by the incident light absorbed by the active area; although other contributions include light energy absorbed by the window aperture and electrical power dissipation of the array. Optical systems must be designed to minimize the light energy falling outside the window clear aperture since any additional thermal load in this area can significantly degrade the reliability of the device.



6.6 Electrical Characteristics

	PARAMETER	TEST CONDITIONS	MIN	TYP MAX	UNIT
V _{OH}	High level output voltage	V _{CC} = 1.8 V, I _{OH} = -2 mA	0.8 x V _{CC}		V
V _{OL}	Low level output voltage	V _{CC} = 1.95 V, I _{OL} = 2 mA		0.2 x V _{CC}	V
I _{OZ}	High impedance output current	V _{CC} = 1.95 V	-40	25	μA
կլ	Low level input current	V _{CC} = 1.95 V, V _I = 0	-1		μA
I _{IH}	High level input current ⁽¹⁾ ⁽²⁾	V_{CC} = 1.95 V, V _I = V _{CC}		110	μA
I _{CC}	Supply current VCC	V _{CC} = 1.95 V		715	mA
IOFFSET	Supply current VOFFSET (2)	V _{OFFSET} = 10.5 V		7	mA
I _{BIAS}	Supply current VBIAS ⁽²⁾ (3)	V _{BIAS} = 18.5 V		2.75	mA
I _{RESET}	Supply current VRESET ⁽³⁾	V _{RESET} = -14.5 V		-5	mA
P _{CC}	Supply power dissipation V_{CC}	V _{CC} = 1.95 V		1394.25	mW
P _{OFFSET}	Supply power dissipation V _{OFFSET} ⁽²⁾	V _{OFFSET} = 10.5 V		73.50	mW
P _{BIAS}	Supply power dissipation V_{BIAS} ^{(2) (3)}	V _{BIAS} = 18.5 V		50.87	mW
P _{RESET}	Supply power dissipation V_{RESET} ⁽³⁾	V _{RESET} = -14.5 V		72.5	mW
P _{TOTAL}	Supply power dissipation V _{TOTAL}			1591.12	mW

Over operating free-air temperature range (unless otherwise noted)

(1) Applies to LVCMOS pins only. Excludes LVDS pins and MBRST (15:0) pins.

(2) To prevent excess current, the supply voltage difference |VBIAS – VOFFSET| must be less than the specified limits listed in the Section 6.4 table.

(3) To prevent excess current, the supply voltage difference |VBIAS – VRESET| must be less than specified limit in *Recommended Operating Conditions*.

6.7 Capacitance at Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
C _{I_lvds}	LVDS input capacitance 2× LVDS	f = 1 MHz			20	pF
CI_nonlvds	Non-LVDS input capacitance 2× LVDS	<i>f</i> = 1 MHz			20	pF
C _{I_tdiode}	Temperature diode input capacitance 2× LVDS	<i>f</i> = 1 MHz			30	pF
Co	Output capacitance	<i>f</i> = 1 MHz			20	pF



6.8 Timing Requirements

			MIN	NOM	MAX	UNIT
SCP ⁽¹⁾						
t _r	Rise time	20% to 80% reference points			30	ns
t _f	Fall time	80% to 20% reference points			30	ns
LVDS ⁽²⁾			· ·			
t _r	Rise slew rate	20% to 80% reference points	0.7	1		V/ns
t _f	Fall slew rate	80% to 20% reference points	0.7	1		V/ns
4		DCLK_C,LVDS pair	2.5			ns
t _C	Clock cycle	DCLK_D, LVDS pair	2.5			ns
+	Pulse duration	DCLK_C LVDS pair	1.19	1.25		ns
t _W	Puise duration	DCLK_D LVDS pair	1.19	1.25		ns
		D_C(15:0) before DCLK_C, LVDS pair	0.275			ns
	Catur time	D_D(15:0) before DCLK_D, LVDS pair	0.275			ns
t _{Su}	Setup time	SCTRL_C before DCLK_C, LVDS pair	0.275			ns
		SCTRL_D before DCLK_D, LVDS pair	0.275			ns
		D_C(15:0) after DCLK_C, LVDS pair	0.195			ns
		D_D(15:0) after DCLK_D, LVDS pair	0.195			ns
t _h	Hold time	SCTRL_C after DCLK_C, LVDS pair	0.195			ns
		SCTRL_D after DCLK_D, LVDS pair	0.195			ns
LVDS ⁽²⁾					I	
t _{SKEW}	Skew time	Channel D relative to Channel C ^{(3) (4)} , LVDS pair	-1.25		1.25	ns

See Figure 6-3 for rise time and fall time for SCP.
 See Figure 6-5 for timing requirements for LVDS.

Channel C (Bus C) includes the following LVDS pairs: DCLK_CN and DCLK_CP, SCTRL_CN and SCTRL_CP, D_CN(15:0) and (3) D_CP(15:0).

(4) Channel D (Bus D) includes the following LVDS pairs: DCLK_DN and DCLK_DP, SCTRL_DN and SCTRL_DP, D_DN(15:0) and D_DP(15:0).

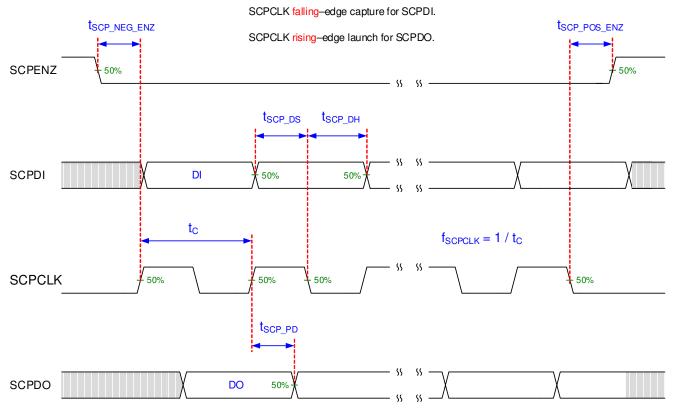


Figure 6-2. SCP Timing Requirements

See *Recommended Operating Conditions* for f_{SCPCLK} , t_{SCP_DS} , t_{SCP_DH} and t_{SCP_PD} specifications.

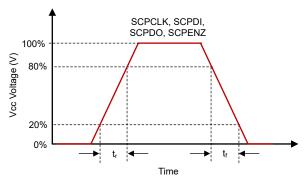


Figure 6-3. SCP Requirements for Rise and Fall

See *Timing Requirements* for t_r and t_f specifications and conditions.

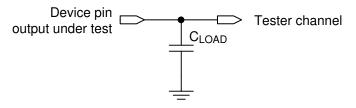
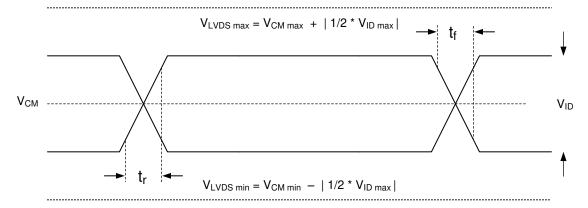


Figure 6-4. Test Load Circuit for Output Propagation Measurement

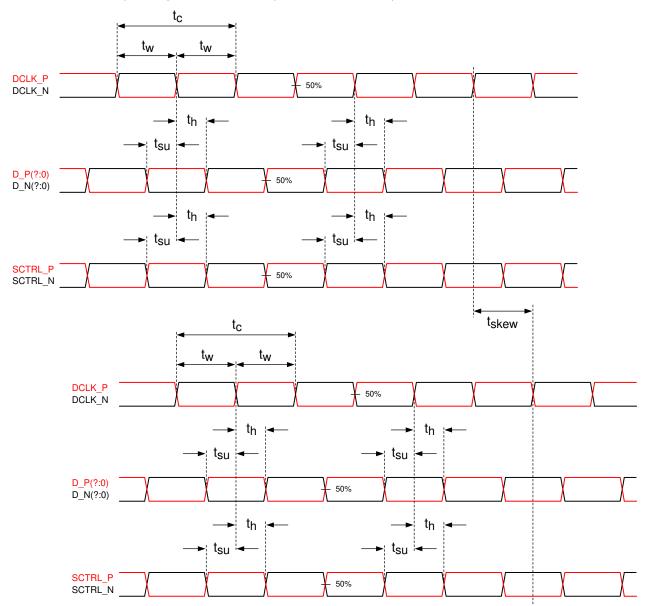
For output timing analysis, the tester pin electronics and its transmission line effects must be taken into account. System designers use IBIS or other simulation tools to correlate the timing reference load to a system environment.







See Recommended Operating Conditions for V_{CM} , V_{ID} , and V_{LVDS} specifications and conditions.







See *Timing Requirements* for timing requirements and LVDS pairs per channel (bus) defining $D_P(?:0)$ and $D_N(?:0)$.

6.9 System Mounting Interface Loads

PARAMETER	MIN	NOM	MAX	UNIT
Thermal interface area ⁽¹⁾			12	kg
Electrical interface area ⁽¹⁾			25	kg

(1) Uniformly distributed within area shown in Figure 6-7

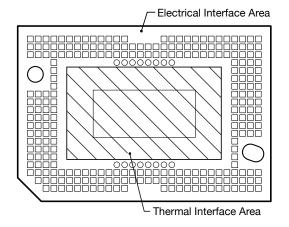


Figure 6-7. System Mounting Interface Loads

6.10 Micromirror Array Physical Characteristics

Table 6-2. Micromirror Array Physical Characteristics

PARAMETER DESCRIPTION			UNIT			
Number of active columns ⁽¹⁾	М	1920	micromirrors			
Number of active rows ⁽¹⁾	N	1200	micromirrors			
Micromirror (pixel) pitch ⁽¹⁾	Р	5.4	μm			
Micromirror active array width ⁽¹⁾	Micromirror Pitch × number of active columns	10.368	mm			
Micromirror active array height ⁽¹⁾	Micromirror Pitch × number of active rows	6.48	mm			
Micromirror active border (Top / Bottom) ⁽²⁾	Pond of micromirrors (POM)	20	micromirrors/side			
Micromirror active border (Right / Left) ⁽²⁾	Pond of micromirrors (POM)	84	micromirrors/side			

(1) See Figure 6-8.

(2) The structure and qualities of the border around the active array includes a band of partially functional micromirrors referred to as the *Pond Of Mirrors* (POM). These micromirrors are structurally and electrically prevented from tilting toward the bright or ON state but still require an electrical bias to tilt toward the OFF state.

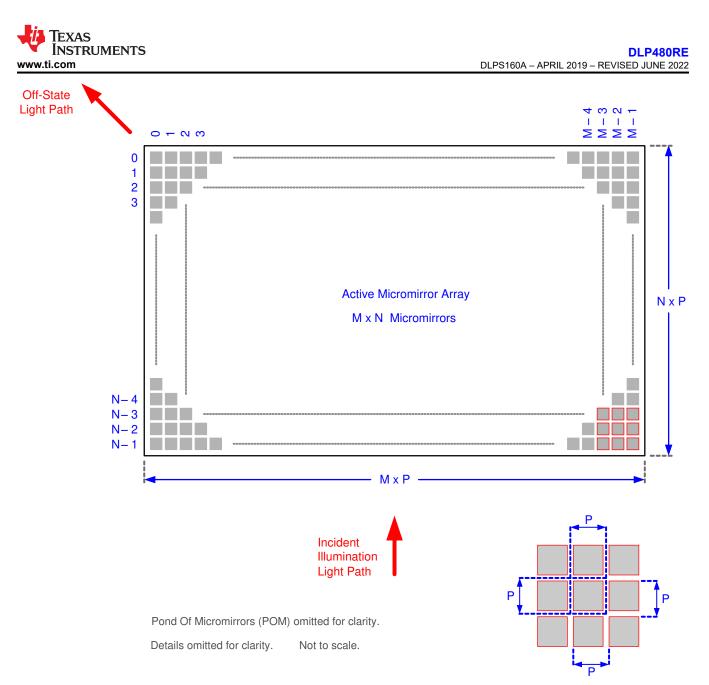


Figure 6-8. Micromirror Array Physical Characteristics



6.11 Micromirror Array Optical Characteristics

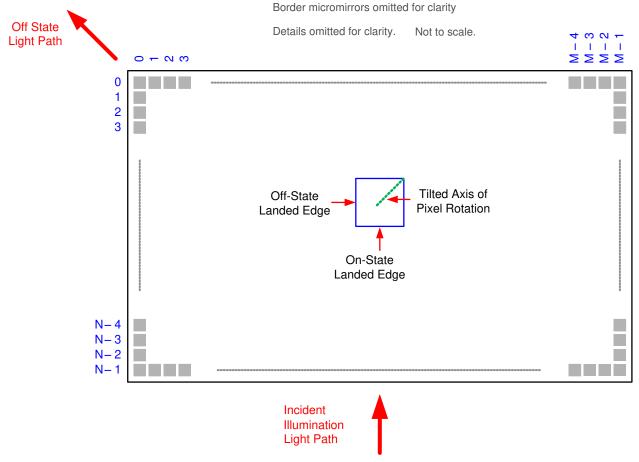
Table 6-3. Micromirror Array Optical Characteristics

PARAMETER		MIN	NOM	MAX	UNIT
Mirror Tilt angle, variation device to device ^{(1) (2)}		15.6	17.0	18.4	degrees
	Adjacent micromirrors			0	
Number of out-of-specification micromirrors ⁽³⁾	Non-Adjacent micromirrors			10	micromirrors

(1) Limits on variability of micromirror tilt angle are critical in the design of the accompanying optical system. Variations in tilt angle within a device may result in apparent non-uniformities, such as line pairing and image mottling, across the projected image. Variations in the average tilt angle between devices may result in colorimetric and system contrast variations.

(2) See Figure 6-9.

(3) An out-of-specification micromirror is defined as a micromirror that is unable to transition between the two landed states within the specified micromirror switching time.



A. Pond of Mirrors (POM) omitted for clarity.

B. Refer to Micromirror Array Physical Characteristics Micromirror Array Physical Characteristics table for M, N, and P specifications.

Figure 6-9. Micromirror Landed Orientation and Tilt



6.12 Window Characteristics

Table 6-4. DMD Window Characteristics

DESCRIPTION	MIN	NOM
Window Material Designation		Corning Eagle XG
Window Refractive Index at 546.1 nm		1.5119
Window Transmittance, minimum within the wavelength range 420–680 nm. Applies to all angles 0–30° AOI. (1) (2)	97%	
Window Transmittance, average over the wavelength range 420–680 nm. Applies to all angles 30–45° AOI. $^{(1)}$	97%	

(1) Single-pass through both surfaces and glass.

(2) Angle of incidence (AOI) is the angle between an incident ray and the normal to a reflecting or refracting surface.

6.13 Chipset Component Usage Specification

Reliable function and operation of the DLP480RE DMD requires that it be used in conjunction with the other components of the applicable DLP chipset, including those components that contain or implement TI DMD control technology. TI DMD control technology is the TI technology and devices for operating or controlling a DLP DMD.



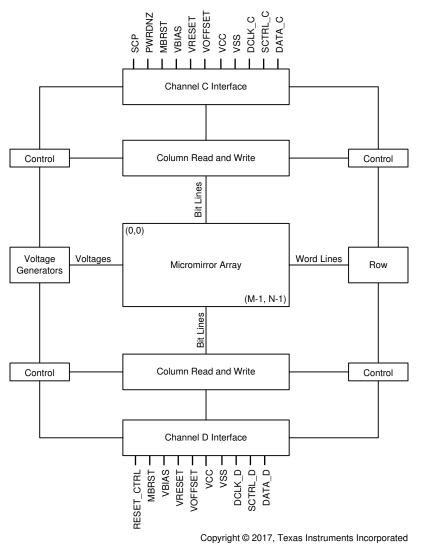
7 Detailed Description

7.1 Overview

The DMD is a 0.48-inch diagonal spatial light modulator which consists of an array of highly reflective aluminum micromirrors. The DMD is an electrical input, optical output micro-optical-electrical-mechanical system (MOEMS). The electrical interface is Low Voltage Differential Signaling (LVDS). The DMD consists of a two-dimensional array of 1-bit CMOS memory cells. The array is organized in a grid of M memory cell columns by N memory cell rows. Refer to the *Function Block Diagram*. The positive or negative deflection angle of the micromirrors can be individually controlled by changing the address voltage of underlying CMOS addressing circuitry and micromirror reset signals (MBRST).

The DLP480RE DMD is part of the chipset comprising the DLP480RE DMD, the DLPC4422 display controller and the DLPA100 power and motor driver. To ensure reliable operation, the DLP480RE DMD must always be used with the DLPC4422 display controller and the DLPA100 power and motor driver.

7.2 Functional Block Diagram



For pin details on Channels C, and D, refer to *Pin Configurations and Functions* and LVDS Interface section of *Section 6.8*. RESET CTRL is utilized in applications when an external reset signal is required.

Figure 7-1. Functional Block Diagram



7.3 Feature Description

7.3.1 Power Interface

The DMD requires 5 DC voltages: DMD_P3P3V, DMD_P1P8V, VOFFSET, VRESET, and VBIAS. DMD_P3P3V is created by the DLPA100 power and motor driver and is used on the DMD board to create the other 4 DMD voltages, as well as powering various peripherals (TMP411, I2C, and TI level translators). DMD_P1P8V is created by the TI PMIC LP38513S and provides the VCC voltage required by the DMD. VOFFSET (10V), VRESET (-14V), and VBIAS(18V) are made by the TI PMIC TPS65145 and are supplied to the DMD to control the micromirrors.

7.3.2 Timing

The data sheet provides timing at the device pin. For output timing analysis, the tester pin electronics and its transmission line effects must be taken into account. Figure 6-4 shows an equivalent test load circuit for the output under test. Timing reference loads are not intended as a precise representation of any particular system environment or depiction of the actual load presented by a production test. System designers use IBIS or other simulation tools to correlate the timing reference load to a system environment. The load capacitance value stated is only for characterization and measurement of AC timing signals. This load capacitance value does not indicate the maximum load the device is capable of driving.

7.4 Device Functional Modes

DMD functional modes are controlled by the DLPC4422 display controller. See the DLPC4422 display controller data sheet or contact a TI applications engineer.

7.5 Optical Interface and System Image Quality Considerations

7.5.1 Optical Interface and System Image Quality

TI assumes no responsibility for end-equipment optical performance. Achieving the desired end-equipment optical performance involves making trade-offs between numerous component and system design parameters. Optimizing system optical performance and image quality strongly relate to optical system design parameter trades. Although it is not possible to anticipate every conceivable application, projector image quality and optical performance is contingent on compliance to the optical system operating conditions described in the following sections.

7.5.1.1 Numerical Aperture and Stray Light Control

The angle defined by the numerical aperture of the illumination and projection optics at the DMD optical areaneeds to be the same. This angle should not exceed the nominal device micromirror tilt angle unless appropriate apertures are added in the illumination and projection pupils to block out flat-state and stray light from the projection lens. The micromirror tilt angle defines DMD capability to separate the "ON" optical path from any other light path, including undesirable flat-state specular reflections from the DMD window, DMD border structures, or other system surfaces near the DMD such as prism or lens surfaces. If the numerical aperture exceeds the micromirror tilt angle, or if the projection numerical aperture angle is more than two degrees larger than the illumination numerical aperture angle, objectionable artifacts may occur in the display border or active arear.

7.5.1.2 Pupil Match

TI's optical and image quality specifications assume that the exit pupil of the illumination optics is nominally centered within 2° of the entrance pupil of the projection optics. Misalignment of pupils can create objectionable artifacts in the display border and active area, which may require additional system apertures to control, especially if the numerical aperture of the system exceeds the pixel tilt angle.

7.5.1.3 Illumination Overfill

The active area of the device is surrounded by an aperture on the inside DMD window surface that masks structures of the DMD chip assembly from normal view, and is sized to anticipate several optical operating conditions. Overfill light illuminating the window aperture can create artifacts from the edge of the window aperture opening and other surface anomalies that may be visible on the screen. Design the illumination optical



system to limit light flux incident anywhere on the window aperture from exceeding approximately 10% of the average flux level in the active area. Depending on the particular system optical architecture, overfill light may have to be further reduced below the suggested 10% level in order to be acceptable.

7.6 Micromirror Array Temperature Calculation

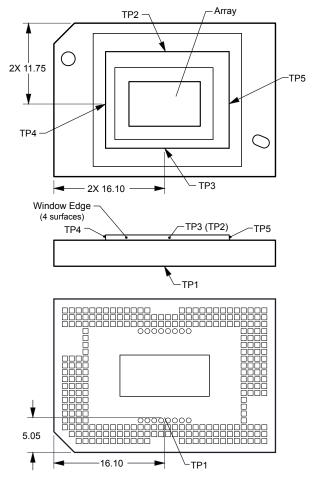


Figure 7-2. DMD Thermal Test Points



Micromirror array temperature can be computed analytically from measurement points on the outside of the package, the package thermal resistance, the electrical power, and the illumination heat load. The relationship between micromirror array temperature and the reference ceramic temperature is provided by the following equations:

 $T_{ARRAY} = T_{CERAMIC} + (Q_{ARRAY} \times R_{ARRAY-TO-CERAMIC})$

 $Q_{ARRAY} = Q_{ELECTRICAL} + (Q_{ILLUMINATION})$

where

- T_{ARRAY} = computed array temperature (°C)
- T_{CERAMIC} = measured ceramic temperature (°C) (TP1 location)
- R_{ARRAY-TO-CERAMIC} = thermal resistance of package from array to ceramic TP1 (°C/Watt)
- Q_{ARRAY} = Total DMD power on the array (Watts) (electrical + absorbed)
- Q_{ELECTRICAL} = nominal electrical power
- $Q_{ILLUMINATION} = (C_{L2W} \times SL)$
- C_{L2W} = Conversion constant for screen lumens to power on DMD (Watts/Lumen)
- SL = measured screen lumens

The electrical power dissipation of the DMD is variable and depends on the voltages, data rates and operating frequencies. A nominal electrical power dissipation to use when calculating array temperature is 0.9 Watts. The absorbed power from the illumination source is variable and depends on the operating state of the micromirrors and the intensity of the light source. The equations shown above are valid for a 1-Chip DMD system with projection efficiency from the DMD to the screen of 87%.

The conversion constant CL2W is based on array characteristics. It assumes a spectral efficiency of 300 lumens/ Watt for the projected light and illumination distribution of 83.7% on the active array, and 16.3% on the array border.

Sample calculations for typical projection application:

 $Q_{ELECTRICAL} = 0.9 W$ $C_{L2W} = 0.00266$ SL = 4000 Im $T_{CERAMIC} = 55.0^{\circ}\text{C}$ $Q_{ARRAY} = 0.9 W + (0.00266 \times 4000 \text{ Im}) = 11.54 W$ $T_{ARRAY} = 55.0^{\circ}\text{C} + (11.54 W \times 0.90^{\circ}\text{C/W}) = 65.39^{\circ}\text{C}$

7.7 Micromirror Landed-On/Landed-Off Duty Cycle

7.7.1 Definition of Micromirror Landed-On/Landed-Off Duty Cycle

The micromirror landed-on/landed-off duty cycle (landed duty cycle) denotes the amount of time (as a percentage) that an individual micromirror is landed in the On state versus the amount of time the same micromirror is landed in the OFF state.

As an example, a landed duty cycle of 100/0 indicates that the referenced pixel is in the ON state 100% of the time (and in the Off state 0% of the time); whereas 0/100 indicate that the pixel is in the OFF state 100% of the time. Likewise, 50/50 indicates that the pixel is ON for 50% of the time (and OFF for 50% of the time).

Note that when assessing landed duty cycle, the time spent switching from one state (ON or OFF) to the other state (OFF or ON) is considered negligible and is thus ignored.

Since a micromirror can only be landed in one state or the other (ON or OFF), the two numbers (percentages) always add to 100.

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7.7.2 Landed Duty Cycle and Useful Life of the DMD

Knowing the long-term average landed duty cycle (of the end product or application) is important because subjecting all (or a portion) of the DMD micromirror array (also called the active array) to an asymmetric landed duty cycle for a prolonged period of time can reduce the DMD usable life.

Note that it is the symmetry or asymmetry of the landed duty cycle that is of relevance. The symmetry of the landed duty cycle is determined by how close the two numbers (percentages) are to being equal. For example, a landed duty cycle of 50/50 is perfectly symmetrical whereas a landed duty cycle of 100/0 or 0/100 is perfectly asymmetrical.

7.7.3 Landed Duty Cycle and Operational DMD Temperature

Operational DMD temperature and landed duty cycle interact to affect DMD usable life, and this interaction can be exploited to reduce the impact that an asymmetrical landed duty cycle has on the DMD usable life. This is quantified in the de-rating curve shown in Figure 6-1. The importance of this curve is that:

- All points along this curve represent the same usable life.
- All points above this curve represent lower usable life (and the further away from the curve, the lower the usable life).
- All points below this curve represent higher usable life (and the further away from the curve, the higher the usable life).

In practice, this curve specifies the Maximum Operating DMD Temperature that the DMD operates at for a given long-term average Landed Duty Cycle.

7.7.4 Estimating the Long-Term Average Landed Duty Cycle of a Product or Application

During a given period of time, the Landed Duty Cycle of a given pixel follows from the image content being displayed by that pixel.

For example, in the simplest case, when displaying pure-white on a given pixel for a given time period, that pixel operates under a 100/0 Landed Duty Cycle during that time period. Likewise, when displaying pure-black, the pixel operates under a 0/100 Landed Duty Cycle.

Between the two extremes (ignoring for the moment color and any image processing that may be applied to an incoming image), the Landed Duty Cycle tracks one-to-one with the gray scale value, as shown in Table 7-1.

LANDED DUTY CYCLE
0/100
10/90
20/80
30/70
40/60
50/50
60/40
70/30
80/20
90/10
100/0

Table 7-1. Grayscale	Value and Landed Duty Cyc	le
----------------------	---------------------------	----



Accounting for color rendition (but still ignoring image processing) requires knowing both the color intensity (from 0% to 100%) for each constituent primary color (red, green, and blue) for the given pixel as well as the color cycle time for each primary color, where "color cycle time" is the total percentage of the frame time that a given primary must be displayed in order to achieve the desired white point.

Use Equation 1 to calculate the landed duty cycle of a given pixel during a given time period

Landed Duty Cycle = (Red_Cycle_% × Red_Scale_Value) + (Green_Cycle_% × Green_Scale_Value) + (Blue_Cycle_% (1) × Blue_Scale_Value)

where

- Red_Cycle_%, represents the percentage of the frame time that red s displayed to achieve the desired white point
- Green_Cycle_% represents the percentage of the frame time that green s displayed to achieve the desired white point
- Blue_Cycle_%, represents the percentage of the frame time that blue is displayed to achieve the desired white point

For example, assume that the red, green and blue color cycle times are 50%, 20%, and 30% respectively (in order to achieve the desired white point), then the Landed Duty Cycle for various combinations of red, green, blue color intensities are as shown in Table 7-2 and Table 7-3.

Table 7-2. Example Landed Duty Cycle for Full-

Color, Color Percentage

CYCLE PERCENTAGE					
RED GREEN BLUE					
50%	20%	30%			

Table 7-3. Example Landed Duty Cycle for Full-Color

S	CALE VALUE		LANDED DUTY
RED	GREEN	BLUE	CYCLE
0%	0%	0%	0/100
100%	0%	0%	50/50
0%	100%	0%	20/80
0%	0%	100%	30/70
12%	0%	0%	6/94
0%	35%	0%	7/93
0%	0%	60%	18/82
100%	100%	0%	70/30
0%	100%	100%	50/50
100%	0%	100%	80/20
12%	35%	0%	13/87
0%	35%	60%	25/75
12%	0%	60%	24/76
100%	100%	100%	100/0



8 Application and Implementation

Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes, as well as validating and testing their design implementation to confirm system functionality.

8.1 Application Information

Texas Instruments DLP[®] technology is a micro-electro-mechanical systems (MEMS) technology that modulates light using a digital micromirror device (DMD). The DMD is a spatial light modulator, which reflects incoming light from an illumination source to one of two directions, towards the projection optics or collection optics. The new TRP pixel with a higher tilt angle increases brightness performance and enables smaller system electronics for size constrained applications. Typical applications using the DLP480RE include business, education, and large venue projectors, interactive displays, and portable smart displays.

The most recent class of chipsets from Texas Instruments is based on a breakthrough micromirror technology, called TRP. With a smaller pixel pitch of 5.4 µm and increased tilt angle of 17 degrees, TRP chipsets enable higher resolution in a smaller form factor and enhanced image processing features while maintaining high optical efficiency. DLP[®] chipsets are a great fit for any system that requires high resolution and high brightness displays.

8.2 Typical Application

The DLP480RE DMD combined with a DLPC4422 digital controller and DLPA100 power management device provides full HD resolution for bright, colorful display applications. A typical display system using the DLP480RE and additional system components is shown in Figure 8-1.

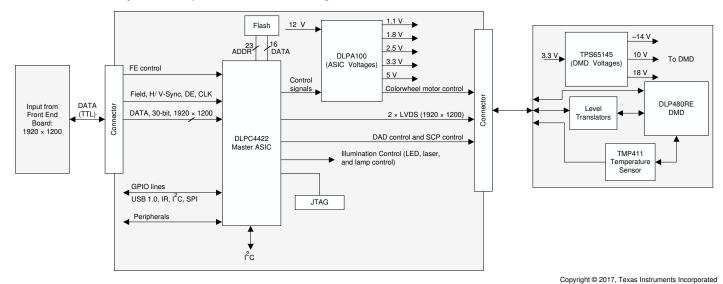


Figure 8-1. Typical WUXGA Application Diagram

8.2.1 Design Requirements

A DLP480RE projection system is created by using the DMD chipset, including the DLP480RE, DLPC4422, and DLPA100. The DLP480RE is used as the core imaging device in the display system and contains a 0.48-inch array of micromirrors. The DLPC4422 controller is the digital interface between the DMD and the rest of the system, taking digital input from front end receiver and driving the DMD over a high speed interface. The DLPA100 power management device provides voltage regulators for the DMD, controller, and illumination functionality.



Other core components of the display system include an illumination source, an optical engine for the illumination and projection optics, other electrical and mechanical components, and software. The illumination source options include lamp, LED, laser or laser phosphor. The type of illumination used and desired brightness will have a major effect on the overall system design and size.

8.2.2 Detailed Design Procedure

For connecting the DLPC4422 display controller and the DLP480RE DMD, see the reference design schematic. For a complete the DLP[®] system, an optical module or light engine is required that contains the DLP480RE DMD, associated illumination sources, optical elements, and necessary mechanical components.

To ensure reliable operation, the DLP480RE DMD must always be used with the DLPC4422 display controllers and a DLPA100 PMIC driver. Refer to PCB Design Requirements for DLP[®] Standard TRP Digital Micromirror Devices for the DMD board design and manufacturing handling of the DMD sub assemblies.

8.2.3 Application Curves

When LED illumination is utilized, typical LED-current-to-Luminance relationship is shown in Figure 8-2

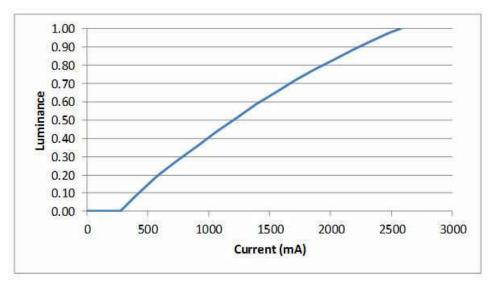


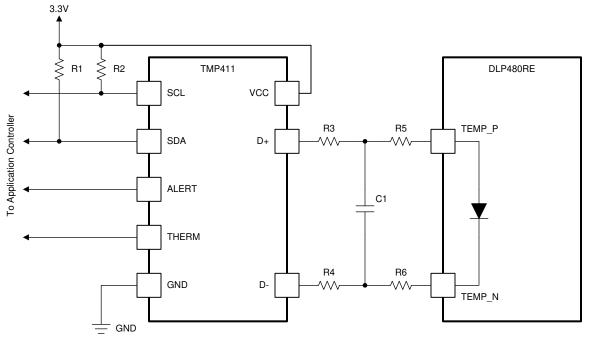
Figure 8-2. Luminance vs. Current

8.3 DMD Die Temperature Sensing

The DMD features a built-in thermal diode that measures the temperature at one corner of the die outside the micromirror array. The thermal diode can be interfaced with the TMP411 temperature sensor as shown in Figure 8-3. The serial bus from the TMP411 can be connected to the DLPC4422 display controller to enable its temperature sensing features. See the DLPC4422 Programmers' Guide for instructions on installing the DLPC4422 controller support firmware bundle and obtaining the temperature readings.

The software application contains functions to configure the TMP411 to read the DMD temperature sensor diode. This data can be leveraged to incorporate additional functionality in the overall system design such as adjusting illumination, fan speeds, and so forth. All communication between the TMP411 and the DLPC4422 controller will be completed using the I²C interface. The TMP411 connects to the DMD via pins B17 and B18 as outlined in *Section 5*.





- A. Details omitted for clarity, see the TI Reference Design for connections to the DLPC4422 controller.
- B. See the TMP411 datasheet for system board layout recommendation.
- C. See the TMP411 datasheet and the TI reference design for suggested component values for R1, R2, R3, R4, and C1.
- D. $R5 = 0 \Omega$. $R6 = 0 \Omega$. Zero ohm resistors need be located close to the DMD package pins.

Figure 8-3. TMP411 Sample Schematic



9 Power Supply Recommendations

The following power supplies are all required to operate the DMD:

- VSS
- VBIAS
- VCC
- VOFFSET
- VRESET

DMD power-up and power-down sequencing is strictly controlled by the DLP® display controller.

CAUTION

For reliable operation of the DMD, the following power supply sequencing requirements must be followed. Failure to adhere to any of the prescribed power-up and power-down requirements may affect device reliability. See Figure 9-1 DMD Power Supply Sequencing Requirements.

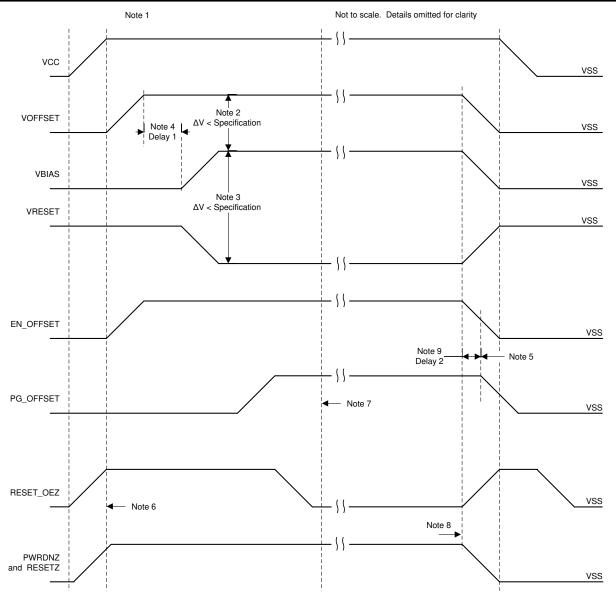
VBIAS, VCC, VOFFSET, and VRESET power supplies must be coordinated during power-up and power-down operations. Failure to meet any of the below requirements will result in a significant reduction in the DMD reliability and lifetime. Common ground VSS must also be connected.

9.1 DMD Power Supply Power-Up Procedure

- During power-up, VCC must always start and settle before VOFFSET plus Delay1 specified in Table 9-1, VBIAS, and VRESET voltages are applied to the DMD.
- During power-up, it is a strict requirement that the voltage difference between VBIAS and VOFFSET must be within the specified limit shown in *Section 6.4*.
- During power-up, there is no requirement for the relative timing of VRESET with respect to VBIAS.
- Power supply slew rates during power-up are flexible, provided that the transient voltage levels follow the requirements specified in *Section 6.1*, in *Section 6.4*, and in Figure 9-1.
- During power-up, LVCMOS input pins must not be driven high until after VCC have settled at operating voltages listed in *Section 6.4*.

9.2 DMD Power Supply Power-Down Procedure

- During power-down, VCC must be supplied until after VBIAS, VRESET, and VOFFSET are discharged to within the specified limit of ground. See Table 9-1.
- During power-down, it is a strict requirement that the voltage difference between VBIAS and VOFFSET must be within the specified limit shown in *Section 6.4*.
- During power-down, there is no requirement for the relative timing of VRESET with respect to VBIAS.
- Power supply slew rates during power-down are flexible, provided that the transient voltage levels follow the requirements specified in *Section 6.1*, in *Section 6.4*, and in Figure 9-1.
- During power-down, LVCMOS input pins must be less than specified in Section 6.4.



- A. See Section 6.4, and .
- B. To prevent excess current, the supply voltage difference |VOFFSET VBIAS| must be less than specified limit in Section 6.4.
- C. To prevent excess current, the supply voltage difference |VBIAS VRESET| must be less than specified limit in Section 6.4.
- D. VBIAS must power up after VOFFSET has powered up, per the Delay1 specification in Table 9-1
- E. PG OFFSET must turn off after EN_OFFSET has turned off, per the Delay2 specification in Table 9-1.
- F. DLP[®] controller software enables the DMD power supplies to turn on after RESET_OEZ is at logic high.
- G. DLP® controller software initiates the global VBIAS command.
- H. After the DMD micromirror park sequence is complete, the DLP[®] controller software initiates a hardware power-down that activates PWRDNZ and disables VBIAS, VRESET and VOFFSET.
- Under power-loss conditions where emergency DMD micromirror park procedures are being enacted by the DLP[®] controller hardware, EN_OFFSET may turn off after PG_OFFSET has turned off. The OEZ signal goes high prior to PG_OFFSET turning off to indicate the DMD micromirror has completed the emergency park procedures.

Figure 9-1. DMD Power Supply Requirements

Table 9-1. DMD Power-Supply Requirements

PARAMETER	DESCRIPTION	MIN	NOM	MAX	UNIT
Delay1	Delay from VOFFSET settled at recommended operating voltage to VBIAS and VRESET power up	1	2		ms



Table 9-1. DMD Power-Supply Requirements (continued)							
PARAMETER	DESCRIPTION	MIN	NOM	MAX	UNIT		
Delay2	PG_OFFSET hold time after EN_OFFSET goes low	100			ns		

10 Layout

10.1 Layout Guidelines

The DLP480RE DMD is part of a chipset that is controlled by the DLPC4422 display controller in conjunction with the DLPA100 power and motor driver. These guidelines are targeted at designing a PCB board with the DLP480RE DMD. The DLP480RE DMD board is a high-speed multi-layer PCB, with primarily high-speed digital logic utilizing dual edge clock rates up to 400MHz for DMD LVDS signals. The remaining traces are comprised of low speed digital LVTTL signals. TI recommends that mini power planes are used for VOFFSET, VRESET, and VBIAS. Solid planes are required for DMD_P3P3V(3.3V), DMD_P1P8V and Ground. The target impedance for the PCB is 50 $\Omega \pm 10\%$ with the LVDS traces being 100 $\Omega \pm 10\%$ differential. TI recommends using an 8 layer stack-up as described in Table 10-1.

10.2 Layout Example

10.2.1 Layers

The layer stack-up and copper weight for each layer is shown in Table 10-1. Small sub-planes are allowed on signal routing layers to connect components to major sub-planes on top or bottom layers if necessary.

LAYER NO.	LAYER NAME	COPPER WT. (oz.)	COMMENTS
1	Side A - DMD only	1.5	DMD, escapes, low frequency signals, power sub-planes.
2	Ground	1	Solid ground plane (net GND).
3	Signal	0.5	50 Ω and 100 Ω differential signals
4	Ground	1	Solid ground plane (net GND)
5	DMD_P3P3V	1	+3.3-V power plane (net DMD_P3P3V)
6	Signal	0.5	50 Ω and 100 Ω differential signals
7	Ground	1	Solid ground plane (net GND).
8	Side B - All other Components	1.5	Discrete components, low frequency signals, power sub-planes

Table 10-1. Layer Stack-Up

10.2.2 Impedance Requirements

TI recommends that the board has matched impedance of 50 $\Omega \pm 10\%$ for all signals. The exceptions are listed in Table 10-2.

Signal Type	Signal Name	Impedance (ohms)						
	DDCP(0:15), DDCN(0:15)							
C channel LVDS differential pairs	DCLKC_P, DCLKC_N	100 ±10% differential across each pair						
	SCTRL_CP, SCTRL_CN							
	DDDP(0:15), DDDN(0:15)							
D channel LVDS differential pairs	DCLKD_P, DCLKD_N	100 ±10% differential across each pair						
	SCTRL_DP, SCTRL_DN	- pan						

Table 10-2. Special Impedance Requirements

10.2.3 Trace Width, Spacing

Unless otherwise specified, TI recommends that all signals follow the 0.005-inch/0.005-inch design rule. Minimum trace clearance from the ground ring around the PWB has a 0.1-inch minimum. An analysis of impedance and stack-up requirements determine the actual trace widths and clearances.

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10.2.3.1 Voltage Signals

Table 10-3. Special Trace Widths, Spacing Requirements

SIGNAL NAME	MINIMUM TRACE WIDTH TO PINS (MIL)	LAYOUT REQUIREMENT
GND	15	Maximize trace width to connecting pin
DMD_P3P3V	15	Maximize trace width to connecting pin
DMD_P1P8V	15	Maximize trace width to connecting pin
VOFFSET	15	Create mini plane from U2 to U3
VRESET	15	Create mini plane from U2 to U3
VBIAS	15	Create mini plane from U2 to U3
All U3 control connections	10	Use 10 mil etch to connect all signals/voltages to DMD pads

11 Device and Documentation Support

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11.2 Device Support

11.2.1 Device Nomenclature

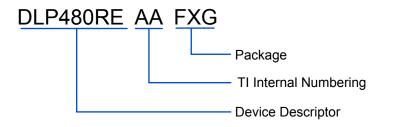


Figure 11-1. Part Number Description

11.2.2 Device Markings

The device marking includes both human-readable information and a 2-dimensional matrix code. The humanreadable information is described in Figure 11-2. The 2-dimensional matrix code is an alpha-numeric character string that contains the DMD part number, part 1 of serial number, and part 2 of serial number. The first character of the DMD Serial Number (part 1) is the manufacturing year. The second character of the DMD Serial Number (part 1) is the manufacturing month. The last character of the DMD Serial Number (part 2) is the bias voltage bin letter.

Example: *1912-513AB GHXXXXX LLLLLLM



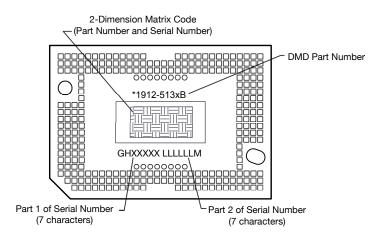


Figure 11-2. DMD Marking Locations

11.3 Documentation Support

11.3.1 Related Documentation

The following documents contain additional information related to the chipset components used with the DLP480RE.

- DLPC4422 Display Controller Data Sheet
- DLPA100 Power and Motor Driver Data Sheet

11.4 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on *Alert me* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

11.5 Support Resources

TI E2E[™] support forums are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

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11.7 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

11.8 Glossary

TI Glossary This glossary lists and explains terms, acronyms, and definitions.

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Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.



PACKAGING INFORMATION

Orderable Device	Status (1)	Package Typ	e Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
DLP480REAAFXG	ACTIVE	CLGA	FXG	257	33	RoHS & Green	Call TI	N / A for Pkg Type			Samples

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

⁽²⁾ RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <= 1000ppm threshold. Antimony trioxide based flame retardants must also meet the <= 1000ppm threshold requirement.

⁽³⁾ MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

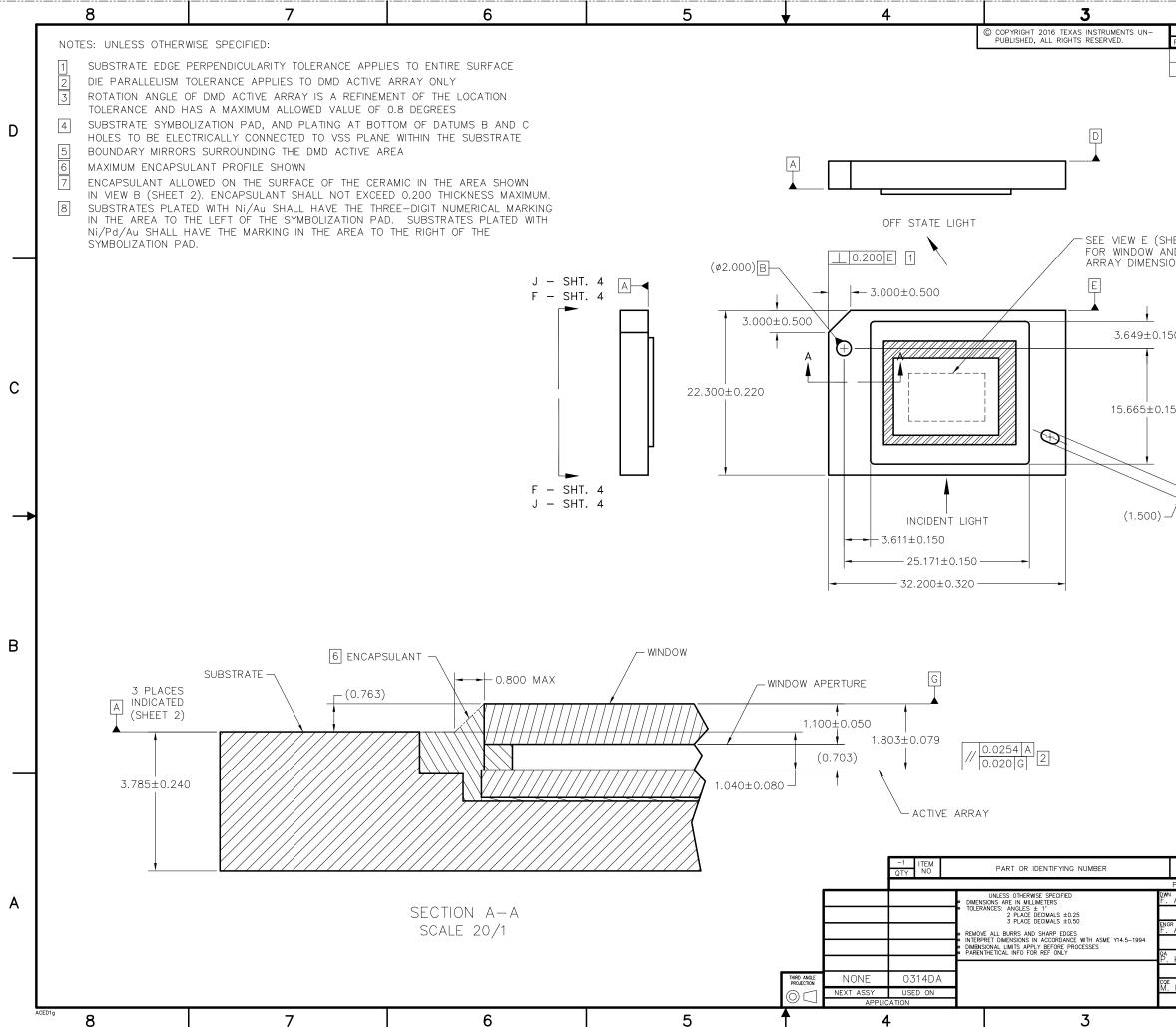
⁽⁴⁾ There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

⁽⁵⁾ Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

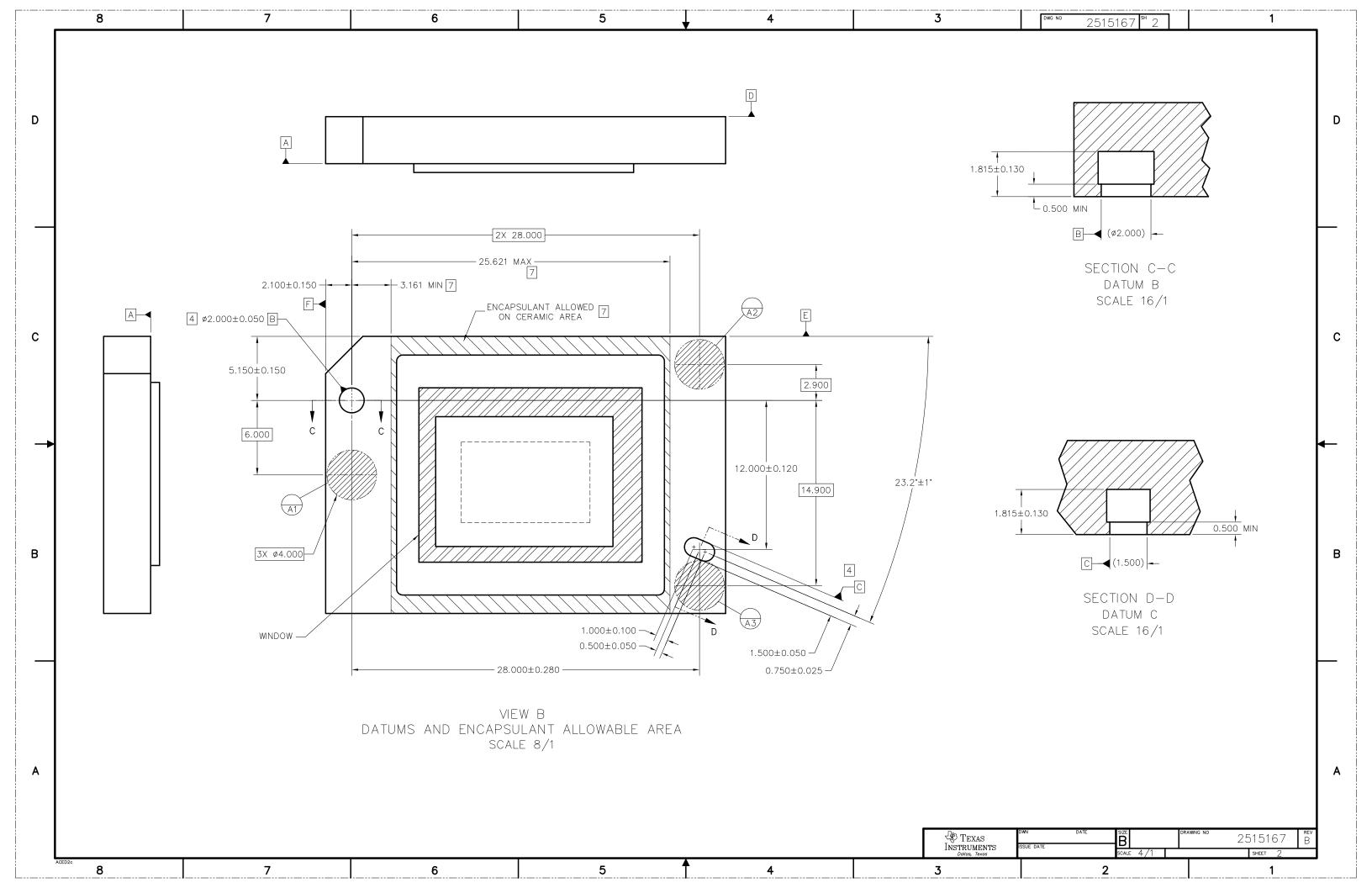
(⁶⁾ Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

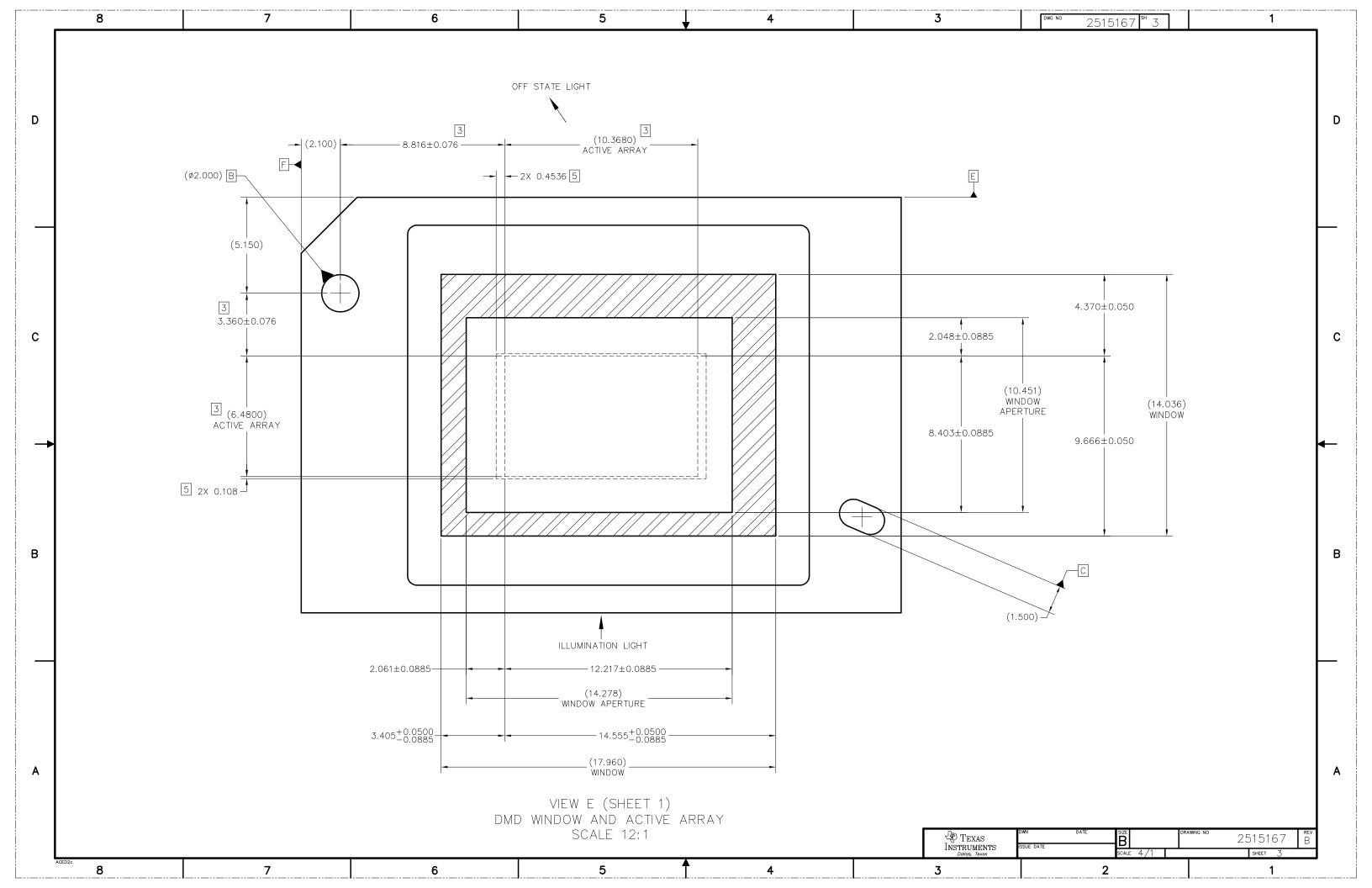
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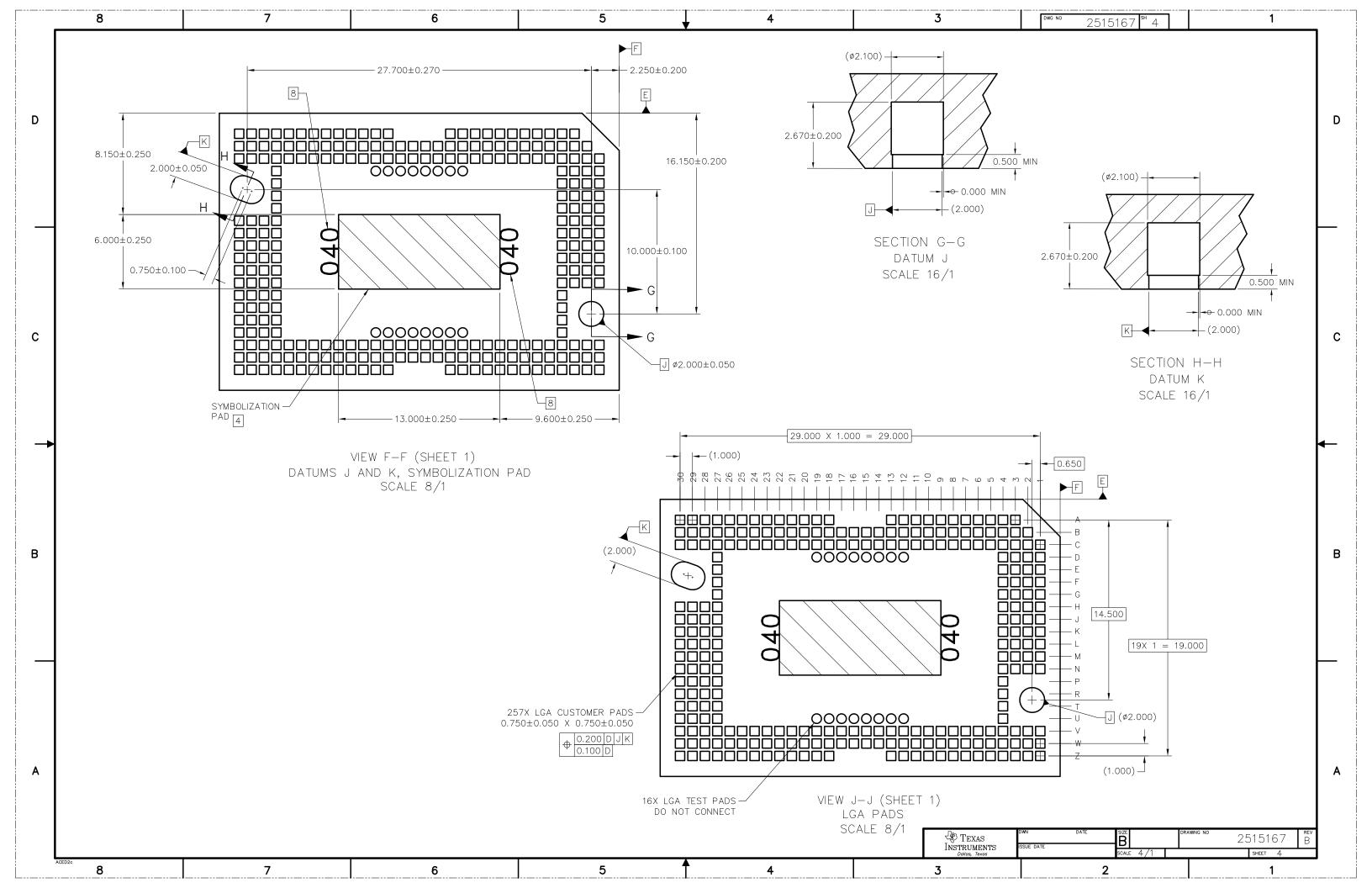
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