













Software

SNVS099I-MAY 1999-REVISED OCTOBER 2016

LM2991

LM2991 Negative Low-Dropout Adjustable Regulator

Features

- Output Voltage Adjustable from -3 V to -24 V, Typically -2 V to -25 V
- Output Current in Excess of 1 A
- Dropout Voltage Typically 0.6 V at 1-A Load
- Low Quiescent Current
- Internal Short Circuit Current Limit
- Internal Thermal Shutdown With Hysteresis
- TTL, CMOS Compatible ON/OFF Switch
- Functional Complement to the LM2941 Series

Applications

- Post Switcher Regulator
- Local, On-Card, Regulation
- **Battery Operated Equipment**
- Industrial
- Instrumentation

3 Description

The LM2991 is a low dropout adjustable negative regulator with a output voltage range between -3 V to -24 V. The LM2991 provides up to 1 A of load current and features a ON /OFF pin for remote shutdown capability.

The LM2991 uses new circuit design techniques to provide a low dropout voltage, low quiescent current and low temperature coefficient precision reference. The dropout voltage at 1-A load current is typically 0.6 V and an ensured worst-case maximum of 1 V over the entire operating temperature range. The quiescent current is typically 1 mA with a 1-A load current and an input-output voltage differential greater than 3 V. A unique circuit design of the internal bias supply limits the quiescent current to only 9 mA (typical) when the regulator is in the dropout mode $(V_{OUT} - V_{IN} \le 3 V)$.

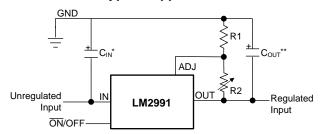
The LM2991 is short-circuit proof, and thermal shutdown includes hysteresis to enhance the reliability of the device when inadvertently overloaded for extended periods. The LM2991 is available in 5lead TO-220 and DDPAK/TO-263 packages and is rated for operation over the automotive temperature range of -40°C to +125°C. Mil-Aero versions are also available.

Device Information⁽¹⁾

PART NUMBER	PACKAGE	BODY SIZE (NOM)		
LM2991	DDPAK/TO-263 (5)	10.20 mm × 9.00 mm		
LIVI2991	TO-220 (5)	14.99 mm × 10.16 mm		

(1) For all available packages, see the orderable addendum at the end of the data sheet.

Typical Application



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 $V_{OUT} = V_{REF} (1 + R2/R1)$

- * Required if the regulator is located further than 6 inches from the power supply filter capacitors. A 1-μF solid tantalum or a 10-µF aluminum electrolytic capacitor is recommended.
- ** Required for stability. Must be at least a 10-μF aluminum electrolytic or a 1-μF solid tantalum to maintain stability. May be increased without bound to maintain regulation during transients. Locate the capacitor as close as possible to the regulator. The equivalent series resistance (ESR) is critical, and should be less than 10Ω over the same operating temperature range as the regulator.



Table of Contents

1	Features	8 Application and Implementation	
2 Applications		8.1 Application Information	
4	Revision History2	9 Power Supply Recommendations	18
5	Pin Configuration and Functions3	10 Layout	16
6	Specifications4	10.1 Layout Guidelines	16
	6.1 Absolute Maximum Ratings 4	10.2 Layout Example	16
	6.2 ESD Ratings	11 Device and Documentation Support	17
	6.3 Recommended Operating Conditions	11.1 Device Support	17
	6.4 Thermal Information	11.2 Related Documentation	17
	6.5 Electrical Characteristics5	11.3 Receiving Notification of Documentation Upda	tes 1
	6.6 Typical Characteristics	11.4 Community Resources	17
7	Detailed Description9	11.5 Trademarks	17
	7.1 Overview 9	11.6 Electrostatic Discharge Caution	18
	7.2 Functional Block Diagram 9	11.7 Glossary	18
	7.3 Feature Description9	12 Mechanical, Packaging, and Orderable	
	7.4 Device Functional Modes	Information	18

4 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

CI	hanges from Revision H (June 2013) to Revision I	Page
•	Added Device Information and ESD Rating tables, Feature Description, Device Functional Modes, Application and Implementation, Power Supply Recommendations, Layout, Device and Documentation Support, and Mechanical, Packaging, and Orderable Information sections; moved some curves to Application Curves section	1
•	Changed footnote 4 of <i>Abs Max</i> table and footnote 1 to <i>Typical Characteristics</i> to eliminate obsolete thermal values for thetaJA; updated values are in <i>Thermal Information</i>	
<u>•</u>	Changed Figure 14 as previous thermal values have been updated	8
CI	hanges from Revision G (April 2013) to Revision H	Page
•	Changed layout of National Semiconductor data sheet to TI format	1



5 Pin Configuration and Functions



Pin Functions

	PIN	1/0	DESCRIPTION		
NO. NAME		1/0	DESCRIPTION		
1	ADJ	1	Feedback pin to the control loop for programming the output voltage.		
2	ON/OFF	I	Logic high enable input		
3	IN	I	Negative Input voltage. Internally connected directly to the thermal tab.		
4	GND	_	Ground		
5	OUT	0	Regulated output voltage		
_	TAB	I	Negative Input voltage. Internally connected directly to the device pin 3. The thermal tab must be connected to a copper area on the PCB at the same potential as device pin 3 (IN) to assure thermal performance, or leave the thermal tab floating. Do NOT connect the thermal tab to any potential other than the same potential at device pin 3. Do NOT connect the thermal tab to ground.		



6 Specifications

6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted) (1)(2)

	MIN	MAX	UNIT
Input voltage	-26	0.3	V
Power dissipation ⁽³⁾	Internal	ly limited	
Storage temperature, T _{stg}	-65	150	°C

- (1) Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) If Military/Aerospace specified devices are required, contact the Texas Instruments Sales Office/Distributors for availability and specifications.
- (3) The maximum power dissipation is a function of $T_{J(MAX)}$, $R_{\theta JA}$, and T_A . The maximum allowable power dissipation at any ambient temperature is $P_D = (T_{J(MAX)} T_A)/R_{\theta JA}$. If this dissipation is exceeded, the die temperature will rise above 125°C, and the LM2991 will eventually go into thermal shutdown at a T_J of approximately 160°C. Refer to *Thermal Shutdown* for more details.

6.2 ESD Ratings

			VALUE	UNIT
V _(ESD)	Electrostatic discharge	Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 (1)	±2000	V

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

6.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted) (1)

	MIN	NOM MAX	UNIT
Junction temperature, T _J	-40	125	°C
ON/OFF pin	0	5	V
Maximum input voltage (operational)	-26		V

⁽¹⁾ Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

6.4 Thermal Information

	THERMAL METRIC ⁽¹⁾	TO-263 (KTT)	TO-220 (NDH) ⁽²⁾	TO-220 (KC) ⁽²⁾	UNIT
		5 PINS	5 PINS	5 PINS	
$R_{\theta JA}^{(3)}$	Junction-to-ambient thermal resistance, High-K	27.8	54.4	56.4	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	41.4	30.1	40.0	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	10.9	33.2	38.6	°C/W
ΨЈТ	Junction-to-top characterization parameter	6.0	11.6	12.8	°C/W
ΨЈВ	Junction-to-board characterization parameter	10.6	36.2	35.3	°C/W
R _θ JC(bot)	Junction-to-case (bottom) thermal resistance	0.7	0.5	0.6	°C/W

- (1) For more information about traditional and new thermal metrics, see Semiconductor and IC Package Thermal Metrics.
- (2) The TO-220 package is vertically mounted in center of a JEDEC High-K test board (JESD 51-7) with no additional heat sink attached. This is a through-hole package; this is NOT a surface-mount package.
- (3) Thermal resistance value R_{0JA} is based on the EIA/JEDEC High-K printed circuit board defined by JESD51-7 High Effective Thermal Conductivity Test Board for Leaded Surface Mount Packages.



6.5 Electrical Characteristics

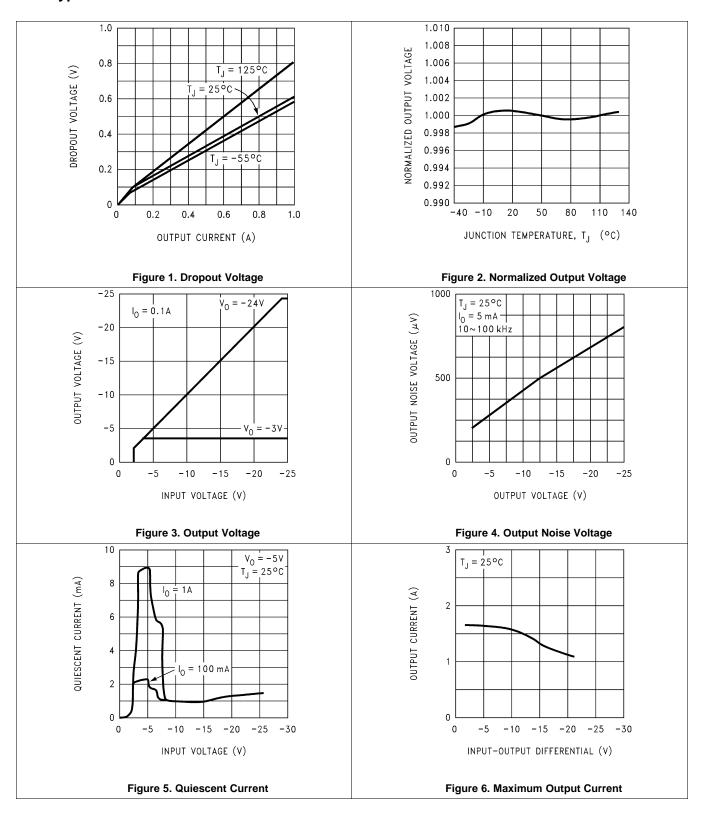
 V_{IN} = -10 V, V_{OUT} = -3 V, I_{OUT} = 1 A, C_{OUT} = 47 μF , R1 = 2.7 $k\Omega$, T_J = 25°C, unless otherwise specified.

$ \begin{array}{l} 5 \text{ mA} \leq I_{OUT} \leq 1 \text{ A} \\ \\ 5 \text{ mA} \leq I_{OUT} \leq 1 \text{ A}, V_{OUT} - 1 \text{ V} \geq V_{IN} > -26 \text{ V} \\ \\ -40^{\circ}\text{C} \leq \text{T}_{J} \leq 125^{\circ}\text{C} \\ \\ \\ \text{Output voltage (V_{OUT})} \\ \\ \text{Uine regulation} \\ \\ \text{Load regulation} \\ \\ \text{IouT} = 5 \text{ mA}, V_{OUT} - 1 \text{ V} > V_{IN} > -26 \text{ V} \\ \\ \text{Load regulation} \\ \\ \text{IouT} = 5 \text{ mA}, V_{OUT} \leq 1 \text{ A} \\ \\ \text{IouT} = 0.1 \text{ A}, \Delta V_{OUT} \leq 100 \text{ mV} \\ \\ \text{IouT} = 0.1 \text{ A}, \Delta V_{OUT} \leq 100 \text{ mV} \\ \\ \text{IouT} = 1 \text{ A}, \Delta V_{OUT} \leq 100 \text{ mV} \\ \\ \text{IouT} = 1 \text{ A}, \Delta V_{OUT} \leq 100 \text{ mV} \\ \\ \text{IouT} = 1 \text{ A}, \Delta V_{OUT} \leq 100 \text{ mV} \\ \\ \text{IouT} = 1 \text{ A}, \Delta V_{OUT} \leq 100 \text{ mV} \\ \\ \text{IouT} = 1 \text{ A}, \Delta V_{OUT} \leq 100 \text{ mV} \\ \\ \text{IouT} = 1 \text{ A}, \Delta V_{OUT} \leq 100 \text{ mV} \\ \\ \text{IouT} = 1 \text{ A}, \Delta V_{OUT} \leq 100 \text{ mV} \\ \\ \text{IouT} = 1 \text{ A}, \Delta V_{OUT} \leq 100 \text{ mV} \\ \\ \text{IouT} = 1 \text{ A}, \Delta V_{OUT} \leq 100 \text{ mV} \\ \\ \text{IouT} = 1 \text{ A}, \Delta V_{OUT} \leq 100 \text{ mV} \\ \\ \text{IouT} = 1 \text{ A}, \Delta V_{OUT} \leq 100 \text{ mV} \\ \\ \text{IouT} = 1 \text{ A}, \Delta V_{OUT} \leq 100 \text{ mV} \\ \\ \text{IouT} = 1 \text{ A}, \Delta V_{OUT} \leq 100 \text{ mV} \\ \\ \text{IouT} = 1 \text{ A}, \Delta V_{OUT} \leq 100 \text{ mV} \\ \\ \text{IouT} = 1 \text{ A}, \Delta V_{OUT} \leq 100 \text{ mV} \\ \\ \text{IouT} = 1 \text{ A}, \Delta V_{OUT} \leq 100 \text{ mV} \\ \\ \text{IouT} = 1 \text{ A}, \Delta V_{OUT} \leq 100 \text{ mV} \\ \\ \text{IouT} = 1 \text{ A}, \Delta V_{OUT} \leq 100 \text{ mV} \\ \\ \text{IouT} = 1 \text{ A}, \Delta V_{OUT} \leq 100 \text{ mV} \\ \\ \text{IouT} = 1 \text{ A}, \Delta V_{OUT} \leq 100 \text{ mV} \\ \\ \text{IouT} = 1 \text{ A}, \Delta V_{OUT} \leq 100 \text{ mV} \\ \\ \text{IouT} = 1 \text{ A}, \Delta V_{OUT} \leq 100 \text{ mV} \\ \\ \text{IouT} = 1 \text{ A}, \Delta V_{OUT} \leq 100 \text{ mV} \\ \\ \text{IouT} = 1 \text{ A}, \Delta V_{OUT} \leq 100 \text{ mV} \\ \\ \text{IouT} = 1 \text{ A}, \Delta V_{OUT} \leq 100 \text{ mV} \\ \\ \text{IouT} = 1 \text{ A}, \Delta V_{OUT} \leq 100 \text{ mV} \\ \\ \text{IouT} = 1 \text{ A}, \Delta V_{OUT} \leq 100 \text{ mV} \\ \\ \text{IouT} = 1 \text{ A}, \Delta V_{OUT} \leq 100 \text{ mV} \\ \\ \text{IouT} = 1 \text{ A}, \Delta V_{OUT} \leq 100 \text{ mV} \\ \\ \text{IouT} = 1 \text{ A}, \Delta V_{OUT} \leq 100 \text{ mV} \\ \\ \text{IouT} = 1 \text{ A}, \Delta V_{OUT} \leq 100 \text{ mV} \\ \\ \text{IouT} = 1 \text{ A}, \Delta V_{OUT} \leq 100 \text{ mV} \\ \\ \text{IouT} = 1 \text{ A}, \Delta V_{OUT} \leq 100 \text{ mV} \\ \\ \text{IouT} = 1 \text{ A}, \Delta V_{OUT} \leq 100 \text{ mV} \\ \\ \text{IouT} = 1 \text{ A}, \Delta V_$	-1.234 -1.27 -24	-1.210 -2	-1.186 -1.15	V	
$-40^{\circ}\text{C} \leq \text{T}_{J} \leq 125^{\circ}\text{C}$ Output voltage (V _{OUT}) $V_{\text{IN}} = -26 \text{ V}$ Line regulation $I_{\text{OUT}} = 5 \text{ mA, V}_{\text{OUT}} - 1 \text{ V} > \text{V}_{\text{IN}} > -26 \text{ V}$ Load regulation $50 \text{ mA} \leq I_{\text{OUT}} \leq 1 \text{ A}$ $I_{\text{OUT}} = 0.1 \text{ A, } \Delta \text{V}_{\text{OUT}} \leq 100 \text{ mV}$ $I_{\text{OUT}} = 0.1 \text{ A, } \Delta \text{V}_{\text{OUT}} \leq 100 \text{ mV}$ $-40^{\circ}\text{C} \leq \text{T}_{J} \leq 125^{\circ}\text{C}$ $I_{\text{OUT}} = 1 \text{ A, } \Delta \text{V}_{\text{OUT}} \leq 100 \text{ mV}$ $I_{\text{OUT}} = 1 \text{ A, } \Delta \text{V}_{\text{OUT}} \leq 100 \text{ mV}$				V	
$\begin{aligned} & V_{\text{IN}} = -26 \text{ V} \\ & \text{Line regulation} & I_{\text{OUT}} = 5 \text{ mA, } V_{\text{OUT}} - 1 \text{ V} > V_{\text{IN}} > -26 \text{ V} \\ & \text{Load regulation} & 50 \text{ mA} \leq I_{\text{OUT}} \leq 1 \text{ A} \\ & I_{\text{OUT}} = 0.1 \text{ A, } \Delta V_{\text{OUT}} \leq 100 \text{ mV} \\ & I_{\text{OUT}} = 0.1 \text{ A, } \Delta V_{\text{OUT}} \leq 100 \text{ mV} \\ & I_{\text{OUT}} = 0.1 \text{ A, } \Delta V_{\text{OUT}} \leq 100 \text{ mV} \\ & I_{\text{OUT}} = 1 \text{ A, } \Delta V_{\text{OUT}} \leq 100 \text{ mV} \end{aligned}$	-24		_3		
$\begin{aligned} & V_{\text{IN}} = -26 \text{ V} \\ & \text{Line regulation} & I_{\text{OUT}} = 5 \text{ mA, } V_{\text{OUT}} - 1 \text{ V} > V_{\text{IN}} > -26 \text{ V} \\ & \text{Load regulation} & 50 \text{ mA} \leq I_{\text{OUT}} \leq 1 \text{ A} \\ & I_{\text{OUT}} = 0.1 \text{ A, } \Delta V_{\text{OUT}} \leq 100 \text{ mV} \\ & I_{\text{OUT}} = 0.1 \text{ A, } \Delta V_{\text{OUT}} \leq 100 \text{ mV} \\ & I_{\text{OUT}} = 0.1 \text{ A, } \Delta V_{\text{OUT}} \leq 100 \text{ mV} \\ & I_{\text{OUT}} = 1 \text{ A, } \Delta V_{\text{OUT}} \leq 100 \text{ mV} \end{aligned}$	-24	25	-3	V	
Load regulation $50 \text{ mA} \le I_{\text{OUT}} \le 1 \text{ A}$ $I_{\text{OUT}} = 0.1 \text{ A}, \Delta V_{\text{OUT}} \le 100 \text{ mV}$ $I_{\text{OUT}} = 0.1 \text{ A}, \Delta V_{\text{OUT}} \le 100 \text{ mV}$ $-40^{\circ}\text{C} \le T_{\text{J}} \le 125^{\circ}\text{C}$ $I_{\text{OUT}} = 1 \text{ A}, \Delta V_{\text{OUT}} \le 100 \text{ mV}$ $I_{\text{OUT}} = 1 \text{ A}, \Delta V_{\text{OUT}} \le 100 \text{ mV}$		-25		V	
Dropout voltage $\begin{aligned} I_{OUT} &= 0.1 \text{ A, } \Delta V_{OUT} \leq 100 \text{ mV} \\ I_{OUT} &= 0.1 \text{ A, } \Delta V_{OUT} \leq 100 \text{ mV} \\ -40^{\circ}\text{C} \leq \text{T}_{\text{J}} \leq 125^{\circ}\text{C} \\ I_{OUT} &= 1 \text{ A, } \Delta V_{OUT} \leq 100 \text{ mV} \\ I_{OUT} &= 1 \text{ A, } \Delta V_{OUT} \leq 100 \text{ mV} \end{aligned}$		0.004	0.04	%/V	
Dropout voltage $\begin{aligned} & I_{OUT} = 0.1 \text{ A, } \Delta V_{OUT} \leq 100 \text{ mV} \\ -40^{\circ}\text{C} \leq T_{\text{J}} \leq 125^{\circ}\text{C} \\ & I_{OUT} = 1 \text{ A, } \Delta V_{OUT} \leq 100 \text{ mV} \\ & I_{OUT} = 1 \text{ A, } \Delta V_{OUT} \leq 100 \text{ mV} \end{aligned}$		0.04%	0.4%		
Dropout voltage $ \frac{-40^{\circ}\text{C} \le \text{T}_{\text{J}} \le 125^{\circ}\text{C}}{\text{I}_{\text{OUT}} = 1 \text{ A}, \Delta \text{V}_{\text{OUT}} \le 100 \text{ mV}} $ $ \text{I}_{\text{OUT}} = 1 \text{ A}, \Delta \text{V}_{\text{OUT}} \le 100 \text{ mV} $		0.1	0.2		
$I_{OUT} = 1 \text{ A, } \Delta V_{OUT} \le 100 \text{ mV}$ $I_{OUT} = 1 \text{ A, } \Delta V_{OUT} \le 100 \text{ mV}$			0.3	V	
$I_{OUT} = 1 \text{ A, } \Delta V_{OUT} \le 100 \text{ mV}$ -40°C \le T _J \le 125°C		0.6	0.8		
	$_{\rm T}$ = 1 A, $\Delta \rm V_{OUT} \le 100~mV$ $_{\rm P}$ $_{\rm T}$		1	V	
Quiescent current I _{OUT} ≤ 1 A		0.7		mA	
$I_{OUT} = 1 \text{ A}, -40^{\circ}\text{C} \le T_{J} \le 125^{\circ}\text{C}$	1 A, −40°C ≤ T _J ≤ 125°C		5	ША	
Dropout quiescent $V_{IN} = V_{OUT}, I_{OUT} \le 1 \text{ A}$ current		16	50	mA	
Ripple rejection $V_{\text{ripple}} = 1 \text{ V}_{\text{RMS}}, \\ f_{\text{ripple}} = 1 \text{ kHz}, I_{\text{OUT}} = 5 \text{ mA}$	50	60		dB	
Output noise 10 Hz to 100 kHz, I _{OUT} = 5 mA		200	450	μV	
V _{OUT} : ON		1.2		V	
ON/OFF input voltage V _{OUT} : ON −40°C ≤ T _J ≤ 125°C			0.8	V	
V _{OUT} : OFF		1.3		V	
V_{OUT} : OFF, -40 °C $\leq T_{J} \leq 125$ °C	2.4				
$\overline{\text{ON}/\text{OFF}}$ input current $\overline{\text{ON}/\text{OFF}} = 0.8 \text{ V}, \text{ V}_{\text{OUT}}$: ON		0.1	10	Δ	
$V_{\overline{ON/OFF}} = 2.4 \text{ V}, V_{\overline{OUT}} : \text{OFF}$		40	100	μΑ	
Output leakage current $V_{IN} = -26 \text{ V}, V_{\overline{ON}/OFF} = 2.4 \text{ V}, V_{OUT} = 0 \text{ V}$					
Current limit V _{OUT} = 0 V		60	250	μΑ	

⁽¹⁾ Typicals are at $T_J = 25^{\circ}C$ and represent the most likely parametric norm.

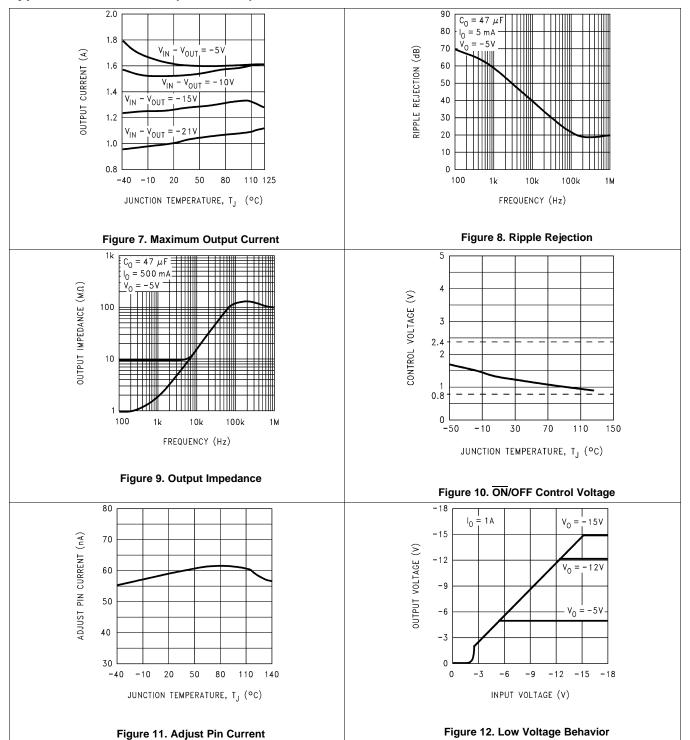
TEXAS INSTRUMENTS

6.6 Typical Characteristics



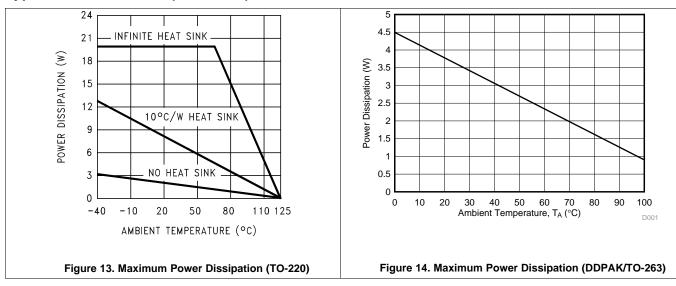


Typical Characteristics (continued)





Typical Characteristics (continued)



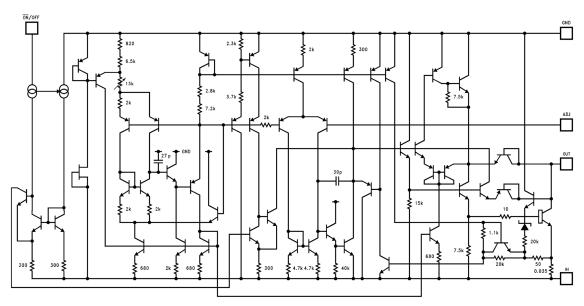


Detailed Description

Overview

The LM2991 is a five-pin, low-dropout, 1-A negative adjustable voltage regulator and negative power supply. ideally suited for a dual-supply system when using together with LM2941 series. The device may also be used as an adjustable current-sink load.

7.2 Functional Block Diagram



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7.3 Feature Description

7.3.1 ON/Off Pin

The LM2991 regulator can be turned off by applying a TTL or CMOS level high signal to the ON/OFF pin. The impedance of the voltage source driving the ON/OFF pin must be low enough to source the ON/OFF pin input current to meet the OFF threshold voltage level, 100 µA maximum at 2.4 V.

If the ON/OFF function is not needed, connect the pin to GND. The ON/OFF pin should not be left floating, as this is not an ensured operating condition. See Figure 15.

7.3.2 Forcing The Output Positive

Due to an internal clamp circuit, the LM2991 can withstand positive voltages on its output. If the voltage source pulling the output positive is DC, the current must be limited to 1.5 A. A current over 1.5 A fed back into the LM2991 could damage the device. The LM2991 output can also withstand fast positive voltage transients up to 26 V, without any current limiting of the source. However, if the transients have a duration of over 1 ms, the output should be clamped with a Schottky diode to ground.

7.3.3 Thermal Shutdown

The LM2991 has an internally set thermal shutdown point of typically 160°C, with approximately 10°C of hysteresis. This thermal shutdown temperature point is outside the specified Recommended Operating Conditions range, above the Absolute Maximum Ratings, and is intended as a safety feature for momentary fault conditions only. Avoid continuous operation near the thermal shutdown temperature as it may have a negative affect on the life of the device.



7.4 Device Functional Modes

7.4.1 Operation with $V_{OUT(TARGET)}$ -5 V \geq $V_{IN} >$ -26 V

The device operates if the input voltage is within $V_{OUT(TARGET)}$ –5 V to –26 V range. At input voltages beyond the V_{IN} requirement, the devices do not operate correctly, and output voltage may not reach target value.



8 Application and Implementation

NOTE

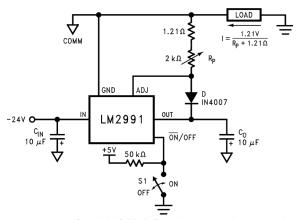
Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

8.1 Application Information

The LM2991 is a 1-A negative adjustable voltage regulator with an operating V_{IN} range of -6 V to -26 V, and a regulated V_{OUT} having 5% accuracy with a maximum rated I_{OUT} current of 1 A. Efficiency is defined by the ratio of output voltage to input voltage because the LM2991 is a linear voltage regulator. To achieve high efficiency, the dropout voltage $(V_{IN} - V_{OUT})$ must be as small as possible, thus requiring a very low dropout LDO.

Successfully implementing an LDO in an application depends on the application requirements. If the requirements are simply input voltage and output voltage, compliance specifications (such as internal power dissipation or stability) must be verified to ensure a solid design. If timing, start-up, noise, PSRR, or any other transient specification is required, the design becomes more challenging.

8.2 Typical Application



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Figure 15. LM2991 Typical Application With Adjustable Current Sink

8.2.1 Design Requirements

For this design example, use the parameters listed in Table 1 as the input parameters.

Table 1. Design Parameters

DESIGN PARAMETER	DESIGN REQUIREMENT
Input voltage	−26 V to −5 V
Output voltage	−2 V to −25 V (typical)
Output current	up to 1 A

8.2.2 Detailed Design Procedure

At 400-mA loading, the dropout of the LM2991 has 1-V maximum dropout over temperature, thus an -5 V headroom is sufficient for operation over both input and output voltage accuracy. The efficiency of the LM2991 in this configuration is V_{OUT} / V_{IN} = 50%.

To achieve the smallest form factor, the TO-263 (KTT) package is selected. Select input and output capacitors in accordance with the *External Capacitors* section. Aluminum capacitances of 470 μ F for the input and 50- μ F capacitors for the output are selected. With an efficiency of 50% and a 400-mA maximum load, the internal power dissipation is 2000 mW, which corresponds to 82.5°C junction temperature rise for the TO-263 package. With an 25°C ambient temperature, the junction temperature is at 107.5°C.

8.2.2.1 External Capacitors

The LM2991 regulator requires an output capacitor to maintain stability. The capacitor must be at least $10-\mu F$ aluminum electrolytic or $1-\mu F$ solid tantalum. The equivalent series resistance (ESR) of the output capacitor must be less than $10~\Omega$, or the zero added to the regulator frequency response by the ESR could reduce the phase margin, creating oscillations. An input capacitor, of at least $1-\mu F$ solid tantalum or $10-\mu F$ aluminum electrolytic, is also needed if the regulator is situated more than 6 inches from the input power supply filter.

8.2.2.1.1 Input Capacitor

TI recommends a solid tantalum or ceramic capacitor whose value is at least 1 μ F, but an aluminum electrolytic (\geq 10 μ F) may be used. However, aluminum electrolytic types should not be used in applications where the ambient temperature can drop below 0°C because their internal impedance increases significantly at cold temperatures.

8.2.2.1.2 Output Capacitor

The output capacitor must meet the ESR limits shown in Figure 16, which means it must have an ESR between about 25 m Ω and 10 Ω .

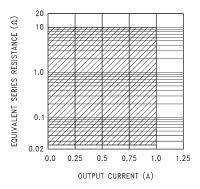


Figure 16. Output Capacitor ESR Range

A solid tantalum (value $\geq 1~\mu\text{F}$) is the best choice for the output capacitor. An aluminum electrolytic ($\geq 10~\mu\text{F}$) may be used if the ESR is in the stable range.

It should be noted that the ESR of a typical aluminum electrolytic will increase by as much as 50x as the temperature is reduced from 25°C down to -40°C, while a tantalum exhibits an ESR increase of about 2x over the same range. For this and other reasons, aluminum electrolytics should not be used in applications where low operating temperatures occur.

The lower stable ESR limit of 25 m Ω means that ceramic capacitors can not be used directly on the output of an LDO. A ceramic ($\geq 2.2~\mu F$) can be used on the output if some external resistance is placed in series with it (1 Ω recommended). Dielectric types X7R or X5R must be used if the temperature range of the application varies more than \pm 25° from ambient to assure the amount of capacitance is sufficient.



8.2.2.2 Ceramic Bypass Capacitors

Many designers place distributed ceramic capacitors whose value is in the range of 1000 pF to 0.1 µF at the power input pins of the IC's across a circuit board. These can cause reduced phase margin or oscillations in LDO regulators.

The advent of multi-layer boards with dedicated power and ground planes has removed the trace inductance that (previously) provided the necessary "de-coupling" to shield the output of the LDO from the effects of bypass capacitors.

Avoid these capacitors, if possible; if ceramic bypass capacitors are used, keep them as far away from the LDO output as is practical.

8.2.2.3 Minimum Load

A minimum load current of 500 μ A is required for proper operation. The external resistor divider can provide the minimum load, with the resistor from the adjust pin to ground set to 2.4 k Ω .

8.2.2.4 Setting The Output Voltage

The output voltage of the LM2991 is set externally by a resistor divider using Equation 1:

$$V_{OUT} = V_{REF} \times (1 + R_2/R_1) - (I_{ADJ} \times R_2)$$

where

•
$$V_{RFF} = -1.21 \text{ V}$$
 (1)

The output voltage can be programmed within the range of -3 V to -24 V, typically an even greater range of -2 V to -25 V. The adjust pin current is about 60 nA, causing a slight error in the output voltage. However, using resistors lower than 100 k Ω makes the error due to the adjust pin current negligible. For example, neglecting the adjust pin current, and setting R2 to 100 k Ω and V_{OUT} to -5 V, results in an output voltage error of only 0.16%.

8.2.2.5 Power Dissipation

Knowing the device power dissipation and proper sizing of the thermal plane connected to the thermal tab is critical to ensuring reliable operation. Device power dissipation depends on input voltage, output voltage, and load conditions and can be calculated with Equation 2.

$$P_{D(MAX)} = (V_{IN(MAX)} - V_{OUT}) \times I_{OUT}$$
(2)

Power dissipation can be minimized, and greater efficiency can be achieved, by using the lowest available voltage drop option that would still be greater than the dropout voltage (V_{DO}). However, keep in mind that higher voltage drops result in better dynamic (that is, PSRR and transient) performance.

Power dissipation and junction temperature are most often related by the junction-to-ambient thermal resistance $(R_{\theta JA})$ of the combined PCB and device package and the temperature of the ambient air (T_A) , according to Equation 3 or Equation 4:

$$T_{J(MAX)} = T_{A(MAX)} + (R_{\theta JA} \times P_{D(MAX)}) \tag{3}$$

$$P_{D(MAX)} = (T_{J(MAX)} - T_{A(MAX)}) / R_{\theta JA}$$
(4)

Unfortunately, this $R_{\theta JA}$ is highly dependent on the heat-spreading capability of the particular PCB design, and therefore varies according to the total copper area, copper weight, and location of the planes. The $R_{\theta JA}$ recorded in *Thermal Information* is determined by the specific EIA/JEDEC JESD51-7 standard for PCB and copper-spreading area, and is to be used only as a relative measure of package thermal performance. For a well-designed thermal layout, $R_{\theta JA}$ is actually the sum of the package junction-to-case (bottom) thermal resistance ($R_{\theta JCbot}$) plus the thermal resistance contribution by the PCB copper area acting as a heat sink.

(5)

(6)



8.2.2.6 Estimating Junction Temperature

The EIA/JEDEC standard recommends the use of psi (Ψ) thermal characteristics to estimate the junction temperatures of surface mount devices on a typical PCB board application. These characteristics are not true thermal resistance values, but rather package specific thermal characteristics that offer practical and relative means of estimating junction temperatures. These psi metrics are determined to be significantly independent of copper-spreading area. The key thermal characteristics $(\Psi_{JT}$ and $\Psi_{JB})$ are given in *Thermal Information* and are used in accordance with Equation 5 or Equation 6.

$$T_{J(MAX)} = T_{TOP} + (\Psi_{JT} \times P_{D(MAX)})$$

where

- P_{D(MAX)} is explained in Equation 4
- T_{TOP} is the temperature measured at the center-top of the device package.

 $T_{J(MAX)} = T_{BOARD} + (\Psi_{JB} \times P_{D(MAX)})$

where

- P_{D(MAX)} is explained in Equation 4
- T_{BOARD} is the PCB surface temperature measured 1-mm from the device package and centered on the package edge.

For more information about the thermal characteristics Ψ_{JT} and Ψ_{JB} , see *Semiconductor and IC Package Thermal Metrics*; for more information about measuring T_{TOP} and T_{BOARD} , see *Using New Thermal Metrics*; and for more information about the EIA/JEDEC JESD51 PCB used for validating $R_{\theta JA}$, see *Thermal Characteristics of Linear and Logic Packages Using JEDEC PCB Designs*. These application notes are available at www.ti.com..

8.2.3 Application Curves

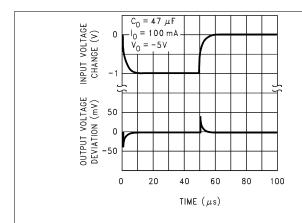


Figure 17. Line Transient Response

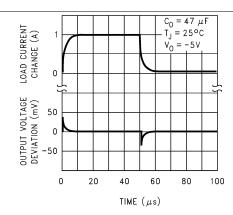


Figure 18. Load Transient Response

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8.2.4 Additional Application Circuits

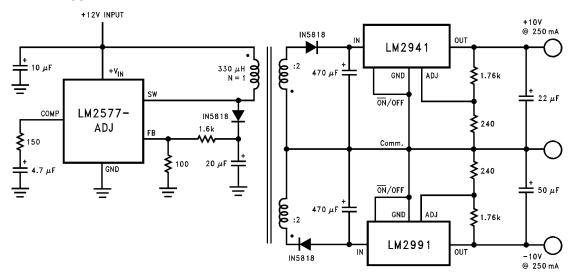


Figure 19. Fully Isolated Post-Switcher Regulator

9 Power Supply Recommendations

The LM2991 is designed to operate from an input voltage supply range between -6 V and -26 V. The input voltage range must provide adequate headroom in order for the device to have a regulated output. This input supply must be well regulated.



10 Layout

10.1 Layout Guidelines

The dynamic performance of the LM2991 is dependent on the layout of the PCB. PCB layout practices that are adequate for typical LDOs may degrade the PSRR, noise, or transient performance of the device. Best performance is achieved by placing C_{IN} and C_{OUT} on the same side of the PCB as the LM2991, and as close as is practical to the package. The ground connections for C_{IN} and C_{OUT} must be back to the LM2991 GND pin using as wide and short of a copper trace as is practical.

Good PC layout practices must be used or instability can be induced because of ground loops and voltage drops. The input and output capacitors must be directly connected to the IN, OUT, and GND pins of the LM2991 using traces which do not have other currents flowing in them (Kelvin connect). The best way to do this is to lay out C_{IN} and C_{OUT} near the device with short traces to the IN, OUT, and GND pins. The regulator ground pin must be connected to the external circuit ground so that the regulator and its capacitors have a single-point ground.

Stability problems have been seen in applications where vias to an internal ground plane were used at the ground points of the LM2991 device and the input and output capacitors. This was caused by varying ground potentials at these nodes resulting from current flowing through the ground plane. Using a single point ground technique for the regulator and its capacitors fixed the problem.

Because high current flows through the traces going into the IN pin and coming from the OUT pin, Kelvin connect the capacitor leads to these pins so there is no voltage drop in series with the input and output capacitors.

10.2 Layout Example

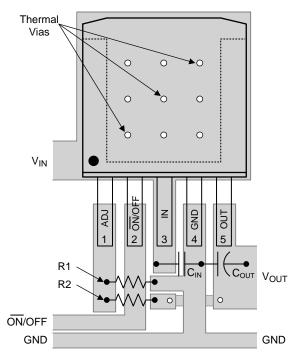


Figure 20. LM2991 TO-263 Board Layout



11 Device and Documentation Support

11.1 Device Support

11.1.1 Device Nomenclature

Dropout Voltage: The input-output voltage differential at which the circuit ceases to regulate against further reduction in input voltage. Measured when the output voltage has dropped 100 mV from the nominal value obtained at (V_{OUT} + 5 V) input, dropout voltage is dependent upon load current and junction temperature.

Input Voltage: The DC voltage applied to the input terminals with respect to ground.

Input-Output Differential: The voltage difference between the unregulated input voltage and the regulated output voltage for which the regulator will operate.

Line Regulation: The change in output voltage for a change in the input voltage. The measurement is made under conditions of low dissipation or by using pulse techniques such that the average chip temperature is not significantly affected.

Load Regulation: The change in output voltage for a change in load current at constant chip temperature.

Long Term Stability: Output voltage stability under accelerated life-test conditions after 1000 hours with maximum rated voltage and junction temperature.

Output Noise Voltage: The rms AC voltage at the output, with constant load and no input ripple, measured over a specified frequency range.

Quiescent Current: That part of the positive input current that does not contribute to the positive load current. The regulator ground lead current.

Ripple Rejection: The ratio of the peak-to-peak input ripple voltage to the peak-to-peak output ripple voltage.

Temperature Stability of V_{OUT}: The percentage change in output voltage for a thermal variation from room temperature to either temperature extreme.

11.2 Related Documentation

For additional information, see the following:

- Semiconductor and IC Package Thermal Metrics
- Using New Thermal Metrics
- Thermal Characteristics of Linear and Logic Packages Using JEDEC PCB Designs

11.3 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on *Alert me* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

11.4 Community Resources

The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

TI E2E™ Online Community TI's Engineer-to-Engineer (E2E) Community. Created to foster collaboration among engineers. At e2e.ti.com, you can ask questions, share knowledge, explore ideas and help solve problems with fellow engineers.

Design Support *TI's Design Support* Quickly find helpful E2E forums along with design support tools and contact information for technical support.

11.5 Trademarks

E2E is a trademark of Texas Instruments.



11.5 Trademarks (continued)

All other trademarks are the property of their respective owners.

11.6 Electrostatic Discharge Caution



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

11.7 Glossary

SLYZ022 — TI Glossary.

This glossary lists and explains terms, acronyms, and definitions.

12 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

Product Folder Links: *LM*2991

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PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
LM2991S	NRND	DDPAK/ TO-263	KTT	5	45	Non-RoHS & Green	Call TI	Level-3-235C-168 HR	-40 to 125	LM2991S P+	
LM2991S/NOPB	ACTIVE	DDPAK/ TO-263	KTT	5	45	RoHS-Exempt & Green	SN	Level-3-245C-168 HR	-40 to 125	LM2991S P+	Samples
LM2991SX	NRND	DDPAK/ TO-263	KTT	5	500	Non-RoHS & Green	Call TI	Level-3-235C-168 HR	-40 to 125	LM2991S P+	
LM2991SX/NOPB	ACTIVE	DDPAK/ TO-263	KTT	5	500	RoHS-Exempt & Green	SN	Level-3-245C-168 HR	-40 to 125	LM2991S P+	Samples
LM2991T	NRND	TO-220	KC	5	45	Non-RoHS & Green	Call TI	Level-1-NA-UNLIM	-40 to 125	LM2991T P+	
LM2991T/LB03	ACTIVE	TO-220	NDH	5	45	Non-RoHS & Green	Call TI	Level-1-NA-UNLIM		LM2991T P+	Samples
LM2991T/LF03	ACTIVE	TO-220	NDH	5	45	RoHS-Exempt & Green	SN	Level-1-NA-UNLIM		LM2991T P+	Samples
LM2991T/NOPB	ACTIVE	TO-220	KC	5	45	RoHS-Exempt & Green	SN	Level-1-NA-UNLIM	-40 to 125	LM2991T P+	Samples

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

⁽²⁾ RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

⁽³⁾ MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

⁽⁴⁾ There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.



PACKAGE OPTION ADDENDUM

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(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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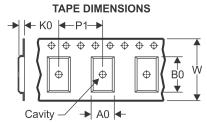
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PACKAGE MATERIALS INFORMATION

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TAPE AND REEL INFORMATION





Α0	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LM2991SX	DDPAK/ TO-263	KTT	5	500	330.0	24.4	10.75	14.85	5.0	16.0	24.0	Q2
LM2991SX/NOPB	DDPAK/ TO-263	KTT	5	500	330.0	24.4	10.75	14.85	5.0	16.0	24.0	Q2

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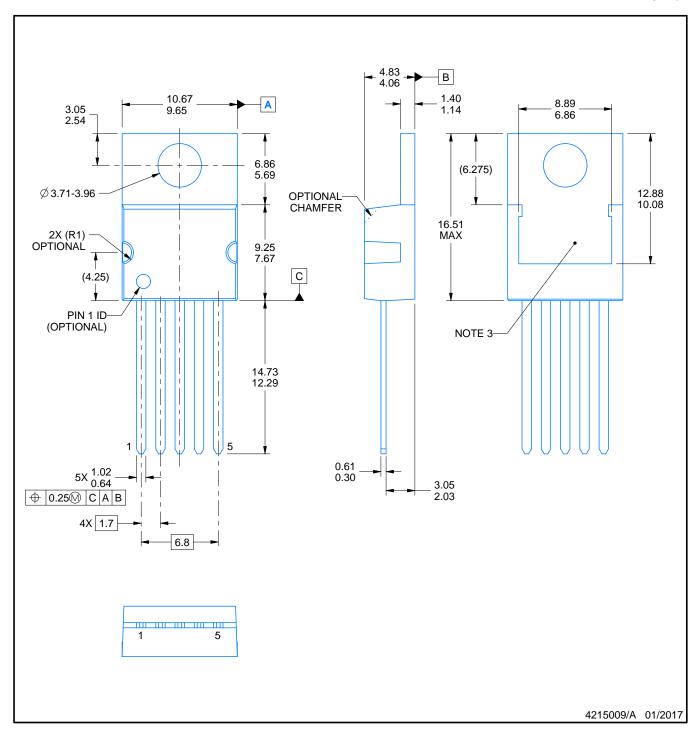


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)	
LM2991SX	DDPAK/TO-263	KTT	5	500	367.0	367.0	45.0	
LM2991SX/NOPB	DDPAK/TO-263	KTT	5	500	367.0	367.0	45.0	



TO-220

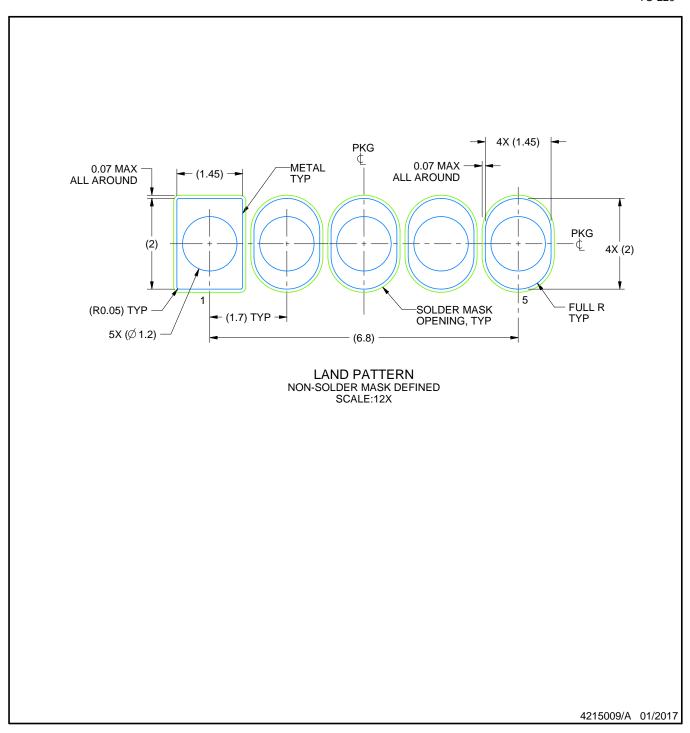


NOTES:

- All controlling linear dimensions are in inches. Dimensions in brackets are in millimeters. Any dimension in brackets or parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
 This drawing is subject to change without notice.
- 3. Shape may vary per different assembly sites.



TO-220







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